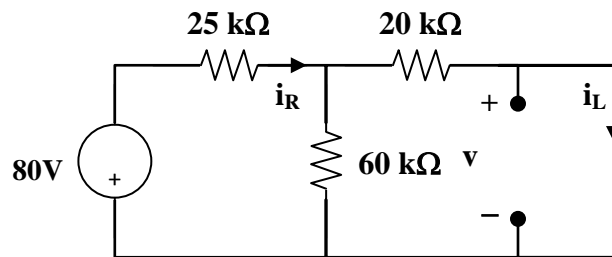


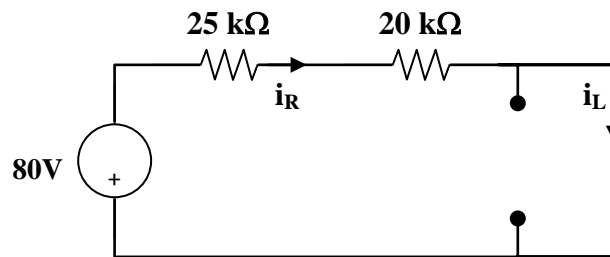
EE 202-Fall 2012(121)
 HW6 - Solution
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Q1

(a) At $t = 0^-$, the equivalent circuit is shown in Figure (a).



(a)



(b)

$$60 \parallel 20 = 15 \text{ kohms}, i_R(0^-) = 80 / (25 + 15) = 2 \text{ mA}.$$

By the current division principle,

$$i_L(0^-) = 60(2 \text{ mA}) / (60 + 20) = 1.5 \text{ mA}$$

$$v_C(0^-) = 0$$

At $t = 0^+$,

$$v_C(0^+) = v_C(0^-) = 0$$

$$i_L(0^+) = i_L(0^-) = \underline{1.5 \text{ mA}}$$

$$80 = i_R(0^+)(25 + 20) + v_C(0^-)$$

$$i_R(0+) = 80/45k = \underline{1.778 \text{ mA}}$$

But,

$$i_R = i_C + i_L$$

$$1.778 = i_C(0+) + 1.5 \text{ or } i_C(0+) = \underline{0.278 \text{ mA}}$$

(b)

$$v_L(0+) = v_C(0+) = 0$$

$$\text{But, } v_L = L di_L/dt \text{ and } di_L(0+)/dt = v_L(0+)/L = 0$$

$$di_L(0+)/dt = \underline{0}$$

$$\text{Again, } 80 = 45i_R + v_C$$

$$0 = 45 di_R/dt + dv_C/dt$$

$$\text{But, } dv_C(0+)/dt = i_C(0+)/C = 0.278 \text{ mA}/1 \mu\text{F} = 278 \text{ V/s}$$

$$\text{Hence, } di_R(0+)/dt = (-1/45)dv_C(0+)/dt = -278/45$$

$$di_R(0+)/dt = \underline{-6.1778 \text{ A/s}}$$

$$\text{Also, } i_R = i_C + i_L$$

$$di_R(0+)/dt = di_C(0+)/dt + di_L(0+)/dt$$

$$-6.1788 = di_C(0+)/dt + 0, \text{ or } di_C(0+)/dt = \underline{-6.1788 \text{ A/s}}$$

(c)

As t approaches infinity, we have the equivalent circuit in Figure (b).

$$i_R(\infty) = i_L(\infty) = 80/45k = \underline{1.778 \text{ mA}}$$

$$i_C(\infty) = C dv(\infty)/dt = \underline{0}.$$

Q2

In the circuit in Fig. 2, calculate $i_o(t)$ and $v_o(t)$ for $t > 0$.

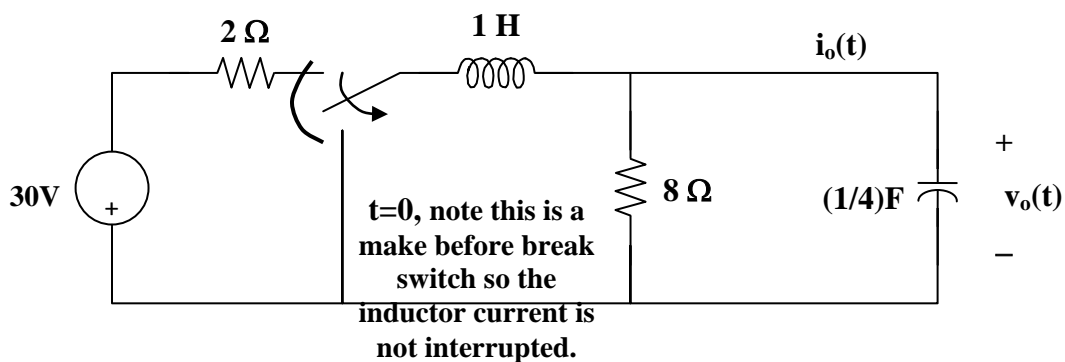


Figure 2 For Problem 2.

At $t = 0^-$, $v_o(0) = (8/(2 + 8))(30) = 24$

For $t > 0$, we have a source-free parallel RLC circuit.

$$\alpha = 1/(2RC) = 1/4$$

$$\omega_o = 1/\sqrt{LC} = 1/\sqrt{1 \times 1/4} = 2$$

Since α is less than ω_o , we have an under-damped response.

$$\omega_d = \sqrt{\omega_o^2 - \alpha^2} = \sqrt{4 - (1/16)} = 1.9843$$

$$v_o(t) = (A_1 \cos \omega_d t + A_2 \sin \omega_d t) e^{-\alpha t}$$

$$v_o(0) = 30(8/(2+8)) = 24 = A_1 \text{ and } i_o(t) = C(dv_o/dt) = 0 \text{ when } t = 0.$$

$$dv_o/dt = -\alpha(A_1 \cos \omega_d t + A_2 \sin \omega_d t) e^{-\alpha t} + (-\omega_d A_1 \sin \omega_d t + \omega_d A_2 \cos \omega_d t) e^{-\alpha t}$$

$$\text{at } t = 0, \text{ we get } dv_o(0)/dt = 0 = -\alpha A_1 + \omega_d A_2$$

$$\text{Thus, } A_2 = (\alpha/\omega_d)A_1 = (1/4)(24)/1.9843 = 3.024$$

$$v_o(t) = \underline{\underline{(24 \cos 1.9843t + 3.024 \sin 1.9843t) e^{-t/4} \text{ volts.}}}$$

$$i_o(t) = Cdv/dt = 0.25[-24(1.9843)\sin 1.9843t + 3.024(1.9843)\cos 1.9843t - 0.25(24 \cos 1.9843t) - 0.25(3.024 \sin 1.9843t)]e^{-t/4}$$

$$= \underline{\underline{[0.000131 \cos 1.9843t - 12.095 \sin 1.9843t] e^{-t/4} \text{ A.}}}$$

Q3

Given that $s_1 = -10$ and $s_2 = -20$, we recall that

$$s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_o^2} = -10, -20$$

$$\text{Clearly, } s_1 + s_2 = -2\alpha = -30 \text{ or } \alpha = 15 = R/(2L) \text{ or } R = 60L \quad (1)$$

$$s_1 = -15 + \sqrt{15^2 - \omega_o^2} = -10 \text{ which leads to } 15^2 - \omega_o^2 = 25$$

$$\text{or } \omega_o = \sqrt{225 - 25} = \sqrt{200} = 1/\sqrt{LC}, \text{ thus } LC = 1/200 \quad (2)$$

Since we have a series RLC circuit, $i_L = i_C = Cdv_C/dt$ which gives,

$$i_L/C = dv_C/dt = [200e^{-20t} - 300e^{-30t}] \text{ or } i_L = 100C[2e^{-20t} - 3e^{-30t}]$$

But, i is also $= 20\{[2e^{-20t} - 3e^{-30t}]\times 10^{-3}\} = 100C[2e^{-20t} - 3e^{-30t}]$

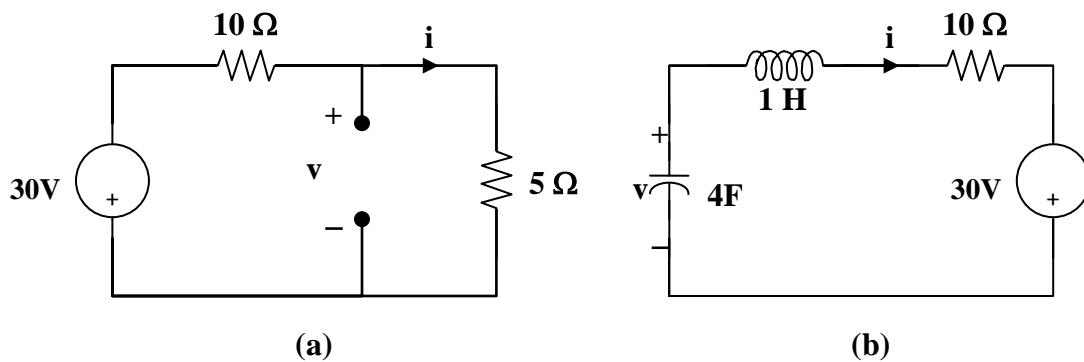
$$\text{Therefore, } C = (0.02/10^2) = \underline{200 \mu\text{F}}$$

$$L = 1/(200C) = \underline{25 \text{ H}}$$

$$R = 30L = \underline{750 \text{ ohms}}$$

Q4

We may transform the current sources to voltage sources. For $t = 0^-$, the equivalent circuit is shown in Figure (a).



$$i(0) = 30/15 = 2 \text{ A}, \quad v(0) = 5 \times 30/15 = 10 \text{ V}$$

For $t > 0$, we have a series RLC circuit.

$$\alpha = R/(2L) = 5/2 = 2.5$$

$$\omega_0 = 1/\sqrt{LC} = 1/\sqrt{4} = 0.25, \text{ clearly } \alpha > \omega_0 \text{ (overdamped response)}$$

$$s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2} = -2.5 \pm \sqrt{6.25 - 0.25} = -4.95, -0.05$$

$$v(t) = V_s + [A_1 e^{-4.95t} + A_2 e^{-0.05t}], \quad v = 20.$$

$$v(0) = 10 = 20 + A_1 + A_2 \tag{1}$$

$$i(0) = Cdv(0)/dt \text{ or } dv(0)/dt = 2/4 = 1/2$$

$$\text{Hence, } \frac{1}{2} = -4.95A_1 - 0.05A_2 \tag{2}$$

From (1) and (2),

$$A_1 = 0, A_2 = -10.$$

$$v(t) = \underline{\{20 - 10e^{-0.05t}\} V}$$

Q5

For $t = 0^-$, we obtain $i(0) = -6/(1 + 2) = -2$ and $v(0) = 2 \times 1 = 2$.

For $t > 0$, the voltage is short-circuited and we have a source-free parallel RLC circuit.

$$\alpha = 1/(2RC) = (1)/(2 \times 1 \times 0.25) = 2$$

$$\omega_o = 1/\sqrt{LC} = 1/\sqrt{1 \times 0.25} = 2$$

Since $\alpha = \omega_o$, we have a critically damped response.

$$s_{1,2} = -2$$

Thus,

$$i(t) = [(A + Bt)e^{-2t}], \quad i(0) = -2 = A$$

$$v = L di/dt = [Be^{-2t}] + [-2(-2 + Bt)e^{-2t}]$$

$$v_o(0) = 2 = B + 4 \quad \text{or} \quad B = -2$$

$$\text{Thus, } i(t) = \underline{\{(-2 - 2t)e^{-2t}\} A}$$

$$\text{and } v(t) = \underline{\{(2 + 4t)e^{-2t}\} V}$$