

3. STATE VARIABLE MODELS (cont.)

ALTERNATIVE SIGNAL-FLOW GRAPH MODELS (CONT.)

Diagonal Form

Consider the transfer function:

$$\frac{Y(s)}{R(s)} = \frac{30(s+1)}{s^3+9s^2+26s+24} = \frac{30(s+1)}{(s+5)(s+2)(s+3)}$$

It is clear that the transient response of the system has three modes, These modes are indicated by the partial fraction expansion as

$$\frac{Y(s)}{R(s)} = \frac{k_1}{(s+5)} + \frac{k_2}{(s+2)} + \frac{k_3}{(s+3)}$$

The coefficients $k_1, k_2,$ and k_3 are called residues and are evaluated by multiplying through by the denominator factor of $\frac{30(s+1)}{(s+5)(s+2)(s+3)}$ corresponding to k_i and setting s equal to the root.

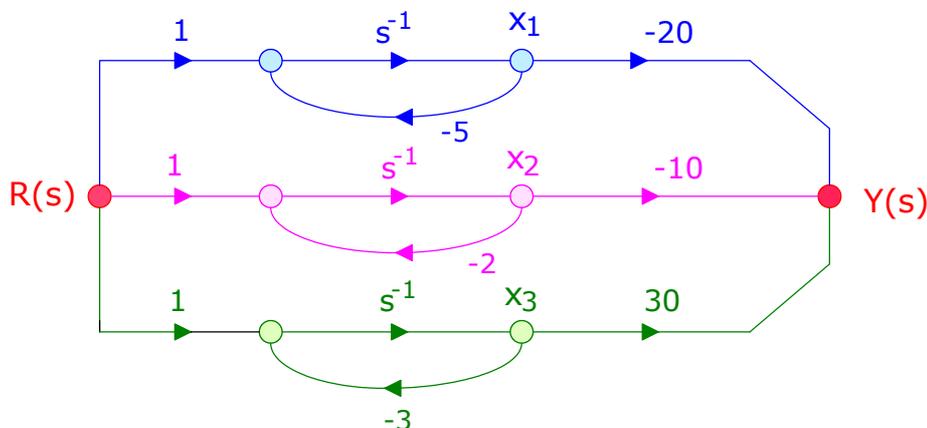
Evaluating $k_1, k_2,$ and k_3 we have

$$k_1 = \left[(s+5) \frac{30(s+1)}{(s+5)(s+2)(s+3)} \right] \Big|_{s=-5} = -20$$

$$k_2 = \left[(s+2) \frac{30(s+1)}{(s+5)(s+2)(s+3)} \right] \Big|_{s=-2} = -10$$

$$k_3 = \left[(s+3) \frac{30(s+1)}{(s+5)(s+2)(s+3)} \right] \Big|_{s=-3} = 30$$

$$\frac{Y(s)}{R(s)} = \frac{-20}{(s+5)} + \frac{-10}{(s+2)} + \frac{30}{(s+3)}$$



Using the above SFG to derive the set of first-order differential equations, we obtain:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -5 & 0 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -3 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix} u(t)$$

$$y = \begin{bmatrix} -20 & -10 & 30 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

THE TRANSFER FUNCTION FROM STATE EQUATIONS

Given the state variable equations, we can obtain the transfer function using a signal-flow graph model and applying Mason's rule. We will now derive a formula for the transfer function of a single-input, single-output system.

Given

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}u \quad ; \quad y = \mathbf{C}\mathbf{x} \quad [D \text{ is assumed} = 0]$$

The Laplace transforms of the above equations are

$$s\mathbf{X}(s) = \mathbf{A}\mathbf{X}(s) + \mathbf{B}U(s) \quad ; \quad Y(s) = \mathbf{C}\mathbf{X}(s)$$

$$(s\mathbf{I} - \mathbf{A})\mathbf{X}(s) = \mathbf{B}U(s)$$

$$\mathbf{X}(s) = (s\mathbf{I} - \mathbf{A})^{-1}\mathbf{B}U(s)$$

$$Y(s) = \mathbf{C}\mathbf{X}(s) = \mathbf{C}(s\mathbf{I} - \mathbf{A})^{-1}\mathbf{B}U(s)$$

$$Y(s) = \mathbf{C}\Phi(s)\mathbf{B}U(s)$$

Note that we do not include initial conditions, since we seek the transfer function.

Therefore the transfer function is $G(s) = \mathbf{C}\Phi(s)\mathbf{B}$

If $D \neq 0$, the transfer function is $G(s) = \mathbf{C}(s\mathbf{I} - \mathbf{A})^{-1}\mathbf{B} + D = \mathbf{C}\Phi(s)\mathbf{B} + D$

Example

Determine the transfer function of the system described by:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & \frac{-1}{C} \\ \frac{1}{L} & \frac{-R}{L} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} \frac{1}{C} \\ 0 \end{bmatrix} u(t) \quad ; \quad y(t) = \begin{bmatrix} 0 & R \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

Solution

$$[sI - A] = \begin{bmatrix} s & \frac{1}{C} \\ -\frac{1}{L} & (s + \frac{R}{L}) \end{bmatrix}; \Delta(s) = |sI - A| = \begin{vmatrix} s & \frac{1}{C} \\ -\frac{1}{L} & (s + \frac{R}{L}) \end{vmatrix} = s^2 + \frac{R}{L}s + \frac{1}{LC}$$

$$\Phi(s) = [sI - A]^{-1} = \begin{bmatrix} s & \frac{1}{C} \\ -\frac{1}{L} & (s + \frac{R}{L}) \end{bmatrix}^{-1} = \frac{1}{\Delta(s)} \begin{bmatrix} (s + \frac{R}{L}) & -\frac{1}{C} \\ \frac{1}{L} & s \end{bmatrix}$$

Then the transfer function is

$$\frac{Y(s)}{U(s)} = [0 \ R] \frac{1}{\Delta(s)} \begin{bmatrix} (s + \frac{R}{L}) & -\frac{1}{C} \\ \frac{1}{L} & s \end{bmatrix} \begin{bmatrix} \frac{1}{C} \\ 0 \end{bmatrix}$$

$$\frac{Y(s)}{U(s)} = \frac{\frac{R}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

EVALUATION OF THE STATE TRANSITION MATRIX

For higher order systems, evaluating $\Phi(s)$ using the formula $\Phi(s) = [sI - A]^{-1}$ is generally inconvenient. The usefulness of the signal-flow graph state model for obtaining the state transition matrix is highlighted.

Consider the system $\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}u$;

The solution for the above system, when $u(t) = 0$, is

$$\mathbf{x}(t) = \Phi(t)\mathbf{x}(0)$$

Taking the Laplace transformation of the above equation, we have

$$\mathbf{X}(s) = \Phi(s)\mathbf{x}(0)$$

Therefore we can evaluate the Laplace transform of the transition matrix from the signal-flow graph by determining the relation between a state variable $X_i(s)$ and the state initial conditions $[x_1(0), x_2(0) \dots x_n(0)]$, using Mason's gain formula.

Thus for a second-order system, we would have

$$X_1(s) = \phi_{11}(s)x_1(0) + \phi_{12}(s)x_2(0)$$

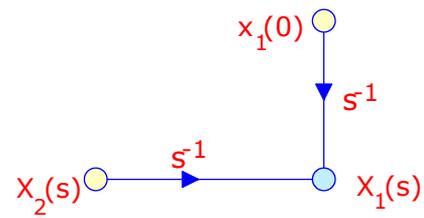
$$X_2(s) = \phi_{21}(s)x_1(0) + \phi_{22}(s)x_2(0)$$

Note that all the elements of the state transition Matrix $\phi_{ij}(s)$, can be obtained by evaluating the individual relationships between $X_i(s)$ and $x_j(0)$ from the state model flow graph.

How to show Initial Conditions on the SFG

Consider the equation $\dot{x}_1 = x_2$; $x_1(0)$
 Taking Laplace transform yields
 $sX_1(s) - x_1(0) = X_2(s)$

The above equation becomes
 $X_1(s) = s^{-1}x_1(0) + s^{-1}X_2(s)$, which is algebraic
 and can be represented by a signal flow graph as shown.



Note that the initial condition of the state x_1 appears as an input to the node representing the state with a branch gain of s^{-1} .

Example

Determine $\Phi(s)$ for the system given by $A = \begin{bmatrix} 0 & -2 \\ 1 & -3 \end{bmatrix}$, $B = \begin{bmatrix} 2 \\ 0 \end{bmatrix}$, $C = [0 \ 3]$
 using two different methods.

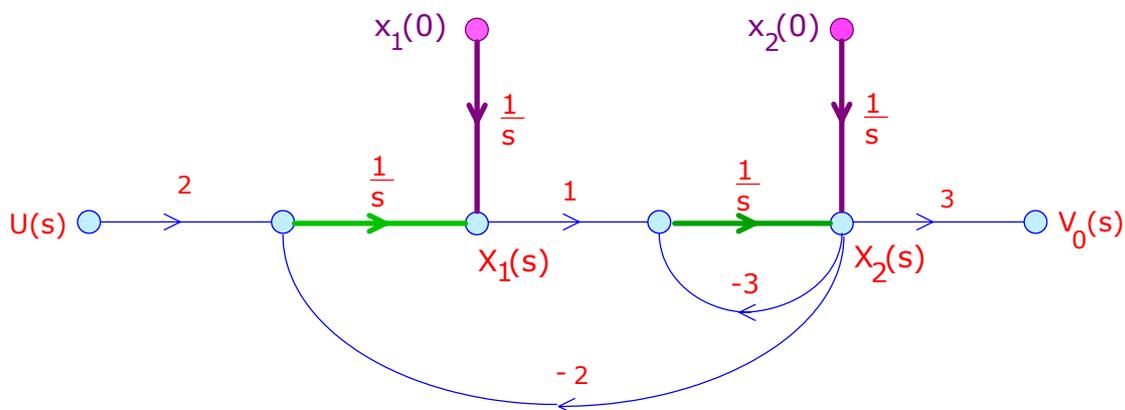
Solution

(1)

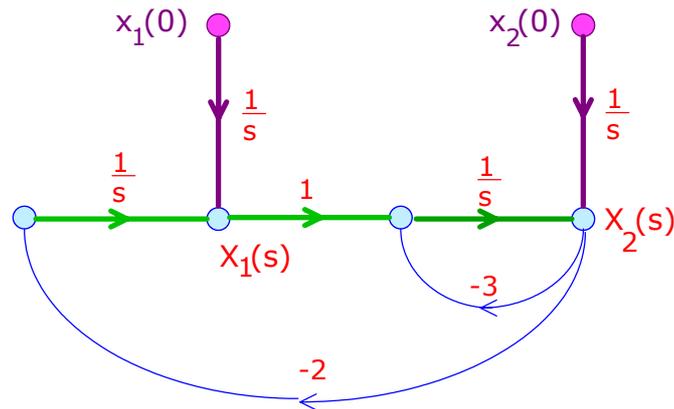
$$[sI - A] = \begin{bmatrix} s & 2 \\ -1 & (s+3) \end{bmatrix}; \Delta(s) = |sI - A| = \begin{vmatrix} s & 2 \\ -1 & (s+3) \end{vmatrix} = s^2 + 3s + 2$$

$$\Phi(s) = [sI - A]^{-1} = \begin{bmatrix} s & 2 \\ -1 & (s+3) \end{bmatrix}^{-1} = \frac{1}{(s^2 + 3s + 2)} \begin{bmatrix} (s+3) & -2 \\ 1 & s \end{bmatrix}$$

(2) Draw a signal-flow graph showing all initial conditions



To obtain $\Phi(s)$, set $U(s) = 0$, and redraw the SFG without the input and output nodes because they are not involved in the evaluation of $\Phi(s)$.



Recall that

$$\begin{aligned} X_1(s) &= \varphi_{11}(s)X_1(0) + \varphi_{12}(s)X_2(0) \\ X_2(s) &= \varphi_{21}(s)X_1(0) + \varphi_{22}(s)X_2(0) \end{aligned}$$

Where $\Phi(s) = \begin{bmatrix} \varphi_{11}(s) & \varphi_{12}(s) \\ \varphi_{21}(s) & \varphi_{22}(s) \end{bmatrix}$

Using Mason's gain formula, we obtain

$$\varphi_{11}(s) = \left. \frac{X_1(s)}{X_1(0)} \right|_{x_2(0)=0} = \frac{\frac{1}{s}(1+3s^{-1})}{1+3s^{-1}+2s^{-2}} = \frac{s+3}{(s^2+3s+2)}$$

$$\varphi_{12}(s) = \left. \frac{X_1(s)}{X_2(0)} \right|_{x_1(0)=0} = \frac{\frac{1}{s}(-2s^{-1})}{1+3s^{-1}+2s^{-2}} = \frac{-2}{(s^2+3s+2)}$$

$$\varphi_{21}(s) = \left. \frac{X_2(s)}{X_1(0)} \right|_{x_2(0)=0} = \frac{\frac{1}{s}(s^{-1})}{1+3s^{-1}+2s^{-2}} = \frac{1}{(s^2+3s+2)}$$

$$\varphi_{22}(s) = \left. \frac{X_2(s)}{X_2(0)} \right|_{x_1(0)=0} = \frac{\frac{1}{s}(1)}{1+3s^{-1}+2s^{-2}} = \frac{s}{(s^2+3s+2)}$$

Hence

$$\Phi(s) = \begin{bmatrix} \frac{s+3}{(s^2+3s+2)} & \frac{-2}{(s^2+3s+2)} \\ \frac{1}{(s^2+3s+2)} & \frac{s}{(s^2+3s+2)} \end{bmatrix}$$

Same answer as (1)

Comments

- We can now find $\Phi(t)$ if we wish

$$\Phi(t) = \mathcal{L}^{-1} \Phi(s) = \begin{bmatrix} (2e^{-t} - e^{-2t}) & (-2e^{-t} + 2e^{-2t}) \\ (e^{-t} - e^{-2t}) & (-e^{-t} + 2e^{-2t}) \end{bmatrix}$$

- We can also find the states and the output for any initial conditions. For example when $x_1(0) = x_2(0) = 1$ and $u(t) = 0$, we have

$$\begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \Phi(t) \begin{bmatrix} x_1(0) \\ x_2(0) \end{bmatrix} = \begin{bmatrix} (2e^{-t} - e^{-2t}) & (-2e^{-t} + 2e^{-2t}) \\ (e^{-t} - e^{-2t}) & (-e^{-t} + 2e^{-2t}) \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \begin{bmatrix} e^{-2t} \\ e^{-2t} \end{bmatrix}$$
