MEAN-SQUARE ANALYSIS OF NORMALIZED LEAKY ADAPTIVE FILTERS.

ALI H. SAYED¹ AND TAREQ Y. AL-NAFFOURI²

¹Electrical Engineering Department University of California, Los Angeles, CA 90095

> ²Electrical Engineering Department Stanford University, CA 94305

ABSTRACT

In this paper, we study leaky adaptive algorithms that employ a general scalar or matrix data nonlinearity. We perform mean-square analysis of this class of algorithms without imposing restrictions on the distribution of the input signal. In particular, we derive conditions on the step-size for stability, and provide closed form expressions for the steady-state performance.

1. ADAPTIVE FILTERING MODEL

In this paper, we consider the following class of leaky adaptive filters:

$$\boldsymbol{w}_{i+1} = (1 - \alpha \mu) \boldsymbol{w}_i + \mu \mathbf{H}(\mathbf{u}_i) \mathbf{u}_i^T \boldsymbol{e}(i) \qquad (1)$$

$$e(i) = d(i) - \mathbf{u}_i \boldsymbol{w}_i \tag{2}$$

$$d(i) = \mathbf{u}_i \boldsymbol{w}^o + v(i) \tag{3}$$

where w_i is an estimate for w^o at iteration i, μ is the step-size, $\alpha \geq 0$ is the leakage parameter, u_i is a row regression vector, v(i) is measurement noise, and $\mathbf{H}(u_i)$ is a matrix data nonlinearity with nonnegative diagonal entries. Usually, $\mathbf{H}(u_i)$ is a multiple of the identity, say $\mathbf{H}(u_i) = \frac{1}{g(u_i)}I$ for some function $g(\cdot)$. Table 1 lists some common examples of data nonlinearities. There are several reasons for incorporating leakage into an adaptive filter update and special cases of (1)-(2) have been studied before in the literature (see, e.g., [1] and [2] and the references therein for motivation and related discussions).

The purpose of this article is to provide a framework for performing mean-square analysis of the general class of leaky algorithms (1)-(2). This is achieved by relying

Table	1:	Exampl	les of	data	non	lineariti	es.

Algorithm	$H(\mathbf{u}_i)$			
NLMS	$\frac{1}{\left\ \mathbf{u}_{i}\right\ ^{2}}I$			
ε-NLMS	$\frac{1}{\epsilon+ \mathbf{u}_i ^2}I$			
sign regressor	$\operatorname{diag}\left(\frac{\operatorname{sign}(u_{i_1})}{u_{i_1}},\ldots,\frac{\operatorname{sign}(u_{i_M})}{u_{i_M}}\right)$			
variable steps	$\operatorname{diag}(\mu_1,\mu_2,\ldots,\mu_M)$			

on the energy-conservation approach developed in [3]-[5]. Among other results, the approach avoids imposing conditions on the statistical distribution of the input sequence (see, e.g., [7, 8]). In addition, the approach enables us to perform both mean-square analysis and transient analysis.

2. DEFINITIONS AND NOTATION

Mean-square analysis of (1)-(2) is carried out in terms of the error quantities:

$$\tilde{\boldsymbol{w}}_i \stackrel{\Delta}{=} \boldsymbol{w}^o - \boldsymbol{w}_i \text{ and } \boldsymbol{e}_a(i) \stackrel{\Delta}{=} \mathbf{u}_i \tilde{\boldsymbol{w}}_i$$
 (4)

and the normalized regressor $\overline{\mathbf{u}}_i = \mathbf{u}_i \mathbf{H}(\mathbf{u}_i)$. These quantities can be used to rewrite the filter relations (1)-(2) as:

$$\tilde{\boldsymbol{w}}_{i+1} = (1 - \alpha \mu) \tilde{\boldsymbol{w}}_i - \mu \overline{\mathbf{u}}_i^T \boldsymbol{e}(i) + \alpha \mu \boldsymbol{w}^o \qquad (5)$$

$$e(i) = e_a(i) + v(i) \tag{6}$$

We shall replace (5) with the more general adaptation

$$\tilde{\boldsymbol{w}}_{i+1} = (1 - \alpha \mu) \tilde{\boldsymbol{w}}_i - \mu \overline{\mathbf{u}}_i^T \boldsymbol{e}(i) + \beta \mu \boldsymbol{w}^o$$
(7)

with separate parameters $\{\alpha, \beta\}$.

We also find it useful to use the compact notation $\|\tilde{\boldsymbol{w}}_i\|_{\Sigma}^2 = \tilde{\boldsymbol{w}}_i^T \boldsymbol{\Sigma} \tilde{\boldsymbol{w}}_i$. This notation is convenient because it enables us to transform operations on $\tilde{\boldsymbol{w}}_i$ into operations on the norm subscript, as demonstrated by the

This work was partially supported by the National Science Foundation under awards ECS-9820765 and CCR-9732376. The work of T. Y. Al-Naffouri was also partially supported by a fellowship from King Fahd University of Petroleum and Minerals, Saudi Arabia.

following properties. Let a_1 and a_2 be scalars and Σ_1 and Σ_2 be symmetric matrices of size M. Then

1) Superposition.

$$a_1 \| \tilde{w}_i \|_{\Sigma_1}^2 + a_2 \| \tilde{w}_i \|_{\Sigma_2}^2 = \| \tilde{w}_i \|_{a_1 \Sigma_1 + a_2 \Sigma_2}^2$$

2) Polarization.

$$(\mathbf{u}_i \boldsymbol{\Sigma}_1 \tilde{\boldsymbol{w}}_i) (\mathbf{u}_i \boldsymbol{\Sigma}_2 \tilde{\boldsymbol{w}}_i) = \| \tilde{\boldsymbol{w}}_i \|_{\boldsymbol{\Sigma}_1 \boldsymbol{u}_i^T \boldsymbol{u}_i \boldsymbol{\Sigma}_2}^2$$

3) Independence. If \tilde{w}_i and u_i are independent,

$$E\left[\left\|\tilde{\boldsymbol{w}}_{\boldsymbol{i}}\right\|_{\Sigma_{1}\boldsymbol{u}_{\boldsymbol{i}}^{T}\boldsymbol{u}_{\boldsymbol{i}}\Sigma_{2}}^{2}\right] = E\left[\left\|\tilde{\boldsymbol{w}}_{\boldsymbol{i}}\right\|_{\Sigma_{1}E\left[\boldsymbol{u}_{\boldsymbol{i}}^{T}\boldsymbol{u}_{\boldsymbol{i}}\right]\Sigma_{2}}^{2}\right]$$

- 4) Linear transformation. For any $N \times M$ matrix A, $\|\mathbf{A}\tilde{w}_i\|_{\Sigma}^2 = \|\tilde{w}_i\|_{A^T \Sigma A}^2$
- 5) Blindness to asymmetry. For any square matrix A,

$$\|\tilde{w}_i\|_A^2 = \|\tilde{w}_i\|_{A^T}^2 = \|\tilde{w}_i\|_{\frac{1}{2}A + \frac{1}{2}A^T}^2$$

6) Notational convention. Using the vector notation, we shall write $\|\tilde{w}_i\|_{\operatorname{vec}(\Sigma_1)}^2 \triangleq \|\tilde{w}_i\|_{\Sigma_1}^2$

The analysis in the sequel relies on the following two assumptions:

- **AN.** The noise sequence v(i) is zero-mean, iid, and is independent of the input regressor u_i .
- AI. The sequence of regressors $\{u_i\}$ is independent with zero mean and autocorrelation matrix \mathbf{R} .

Observe that we are not requiring the input to be Gaussian.

3. MEAN-SQUARE PERFORMANCE

To study the mean-square performance of the leaky adaptive filters, we need to develop a recursion for the weight-error energy. We therefore start with recursion (7) and compute the energies of both sides to arrive at, after taking expectations,

$$E\left[\|\tilde{\boldsymbol{w}}_{i+1}\|_{\Sigma}^{2}\right] = (1 - \alpha \mu)^{2} E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{\Sigma}^{2}\right]$$
$$-2\mu(1 - \alpha \mu) E\left[\tilde{\boldsymbol{w}}_{i}^{T} \mathbf{u}_{i}^{T} \overline{\mathbf{u}}_{i} \Sigma \tilde{\boldsymbol{w}}_{i}\right] + \mu^{2} E\left[e_{a}^{2}(i)\|\overline{\mathbf{u}}_{i}\|_{\Sigma}^{2}\right]$$
$$+2\mu\beta(1 - \alpha\mu) E\left[\boldsymbol{w}^{\sigma^{T}} \Sigma \tilde{\boldsymbol{w}}_{i}\right] - 2\mu^{2}\beta E\left[\boldsymbol{w}^{\sigma^{T}} \mathbf{u}_{i}^{T} \overline{\mathbf{u}}_{i} \Sigma \tilde{\boldsymbol{w}}_{i}\right]$$
$$+\mu^{2} \sigma_{v}^{2} E\left[\|\overline{\mathbf{u}}_{i}\|_{\Sigma}^{2}\right] + \mu^{2} \beta^{2} \|\boldsymbol{w}^{o}\|_{\Sigma}^{2}$$
(8)

In the above calculation, we used assumption AN to eliminate three noise cross-terms. The above recursion

can be expressed more compactly by using the polarization and asymmetry properties, in addition to the independence assumption, to write

$$E\left[\tilde{\boldsymbol{w}}_{i}^{T}\boldsymbol{u}_{i}^{T}\overline{\boldsymbol{u}}_{i}\Sigma\tilde{\boldsymbol{w}}_{i}\right] = E\left[\left\|\tilde{\boldsymbol{w}}_{i}\right\|_{\frac{1}{2}E\left[\boldsymbol{u}_{i}^{T}\overline{\boldsymbol{u}}_{i}\right]\Sigma+\frac{1}{2}\Sigma E\left[\overline{\boldsymbol{u}}_{i}^{T}\boldsymbol{u}_{i}\right]\right] (9)$$
$$E\left[e_{a}^{2}(i)\left\|\overline{\boldsymbol{u}}_{i}\right\|_{\Sigma}^{2}\right] = E\left[\left\|\tilde{\boldsymbol{w}}_{i}\right\|_{E\left[\left\|\overline{\boldsymbol{u}}_{i}\right\|_{\Sigma}^{2}\boldsymbol{u}_{i}^{T}\boldsymbol{u}_{i}\right]\right] (10)$$

and

$$(1 - \alpha \mu) E\left[\boldsymbol{w}^{o^{T}} \boldsymbol{\Sigma} \tilde{\boldsymbol{w}}_{i}\right] - \mu E\left[\boldsymbol{w}^{o^{T}} \mathbf{u}_{i}^{T} \overline{\mathbf{u}}_{i} \boldsymbol{\Sigma} \boldsymbol{w}_{i}\right] = \boldsymbol{w}^{o^{T}} \boldsymbol{\Sigma} \boldsymbol{J} E\left[\tilde{\boldsymbol{w}}_{i}\right] \quad (11)$$

where we defined

$$\mathbf{J} \stackrel{\Delta}{=} E\left[\mathbf{I} - \boldsymbol{\mu} \mathcal{U}_{i}^{T}\right], \qquad \mathcal{U}_{i} \stackrel{\Delta}{=} \alpha \mathbf{I} + \mathbf{u}_{i}^{T} \overline{\mathbf{u}}_{i}$$
(12)

Substituting (9)-(11) into (8), yields

$$E\left[\|\tilde{\boldsymbol{w}}_{i+1}\|_{\Sigma}^{2}\right] = E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{\Sigma'}^{2}\right] + \mu^{2}\sigma_{v}^{2}E\left[\|\bar{\mathbf{u}}_{i}\|_{\Sigma}^{2}\right] + \mu^{2}\beta^{2}\|\boldsymbol{w}^{o}\|_{\Sigma}^{2} + 2\mu\beta\boldsymbol{w}^{o^{T}}\boldsymbol{\Sigma}JE\left[\bar{\boldsymbol{w}}_{i}\right]$$
(13)

where Σ' is related to Σ via

$$\boldsymbol{\Sigma}' = (1 - \alpha \mu)^2 \boldsymbol{\Sigma} - \mu (1 - \alpha \mu) \boldsymbol{\Sigma} \boldsymbol{E} \left[\mathbf{\bar{u}}_i^T \mathbf{u}_i \right]$$
$$-\mu (1 - \alpha \mu) \boldsymbol{E} \left[\mathbf{u}_i^T \mathbf{\bar{u}}_i \right] \boldsymbol{\Sigma} + \mu^2 \boldsymbol{E} [\| \mathbf{\bar{u}}_i \|_{\Sigma}^2 \mathbf{u}_i^T \mathbf{u}_i]$$
(14)

Relations (13)-(14) (or, equivalently, relations (16) -(17) below and ultimately (19)) can be used to characterize the mean-square performance of the adaptive filter. In particular, they can be used to derive conditions for mean-square stability, as well as expressions for the steady-state mean-square error and mean-square deviation of an adaptive filter. To this end, note that the above recursion for Σ can be rewritten more compactly, using the vec operation and the Kronecker product notation, as

$$\sigma' = F\sigma$$
(15)

where $\boldsymbol{\sigma} = \operatorname{vec}(\boldsymbol{\Sigma})$, $\boldsymbol{\sigma}' = \operatorname{vec}(\boldsymbol{\Sigma}')$, and

$$\mathbf{F} = E\left[(\mathbf{I} - \mu \mathcal{U}_i) \otimes (\mathbf{I} - \mu \mathcal{U}_i)\right]$$
(16)

In light of (15), recursion (13) becomes

$$\frac{E\left[\left\|\tilde{\boldsymbol{w}}_{i+1}\right\|_{\sigma}^{2}\right] = E\left[\left\|\tilde{\boldsymbol{w}}_{i}\right\|_{F\sigma}^{2}\right] + \mu^{2}\sigma_{v}^{2}E\left[\left\|\overline{\mathbf{u}}_{i}\right\|_{\sigma}^{2}\right] + \mu^{2}\beta^{2}\left\|\boldsymbol{w}^{o}\right\|_{\sigma}^{2} + 2\mu\beta\boldsymbol{w}^{o^{T}}\boldsymbol{\Sigma}\boldsymbol{J}E\left[\tilde{\boldsymbol{w}}_{i}\right] - \mu^{2}\beta^{2}\left\|\boldsymbol{w}^{o}\right\|_{\sigma}^{2} + \mu^{2}\left\|\boldsymbol{w}^{o}\right\|_{\sigma}^{2} + \mu^{2}\left\|$$

To make this recursion self-contained, we need a recursion for $E[\bar{w}_i]$, which can be obtained by evaluating the expected value of both sides of (7):

$$E\left[\tilde{\boldsymbol{w}}_{i}\right] = \boldsymbol{J}E\left[\tilde{\boldsymbol{w}}_{i-1}\right] + \mu\beta\boldsymbol{w}^{o}$$
(18)

Recursion (18) is what we need to supplement (17) and produce the desired self-contained relation. To this end, let us write (17) explicitly for $\{\sigma, F\sigma, \dots, F^{M^2-1}\sigma\}$:

$$\begin{cases} E\left[\|\tilde{\boldsymbol{w}}_{i+1}\|_{\sigma}^{2}\right] &= E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{F\sigma}^{2}\right] + \mu^{2}\sigma_{v}^{2}E\left[\|\overline{\mathbf{u}}_{i}\|_{\sigma}^{2}\right] \\ + \mu^{2}\beta^{2}\|\boldsymbol{w}^{\circ}\|_{\sigma}^{2} + 2\mu\beta\boldsymbol{f}_{0}^{T}E\left[\bar{\boldsymbol{w}}_{i}\right] \\ E\left[\|\tilde{\boldsymbol{w}}_{i+1}\|_{F\sigma}^{2}\right] &= E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{F\sigma}^{2}\right] + \mu^{2}\sigma_{v}^{2}E\left[\|\overline{\mathbf{u}}_{i}\|_{F\sigma}^{2}\right] \\ + \mu^{2}\beta^{2}\|\boldsymbol{w}^{\circ}\|_{F\sigma}^{2} + 2\mu\beta\boldsymbol{f}_{1}^{T}E\left[\bar{\boldsymbol{w}}_{i}\right] \\ \vdots \\ E\left[\|\tilde{\boldsymbol{w}}_{i+1}\|_{FM^{2}-1\sigma}^{2}\right] &= E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{FM^{2}\sigma}^{2}\right] \\ + \mu^{2}\sigma_{v}^{2}E\left[\|\overline{\mathbf{u}}_{i}\|_{FM^{2}-1\sigma}^{2}\right] \\ + \mu^{2}\beta^{2}\|\boldsymbol{w}^{\circ}\|_{FM^{2}-1\sigma}^{2} \\ + 2\mu\beta\boldsymbol{f}_{M^{2}-1}^{T}E\left[\bar{\boldsymbol{w}}_{i}\right] \\ &= -p_{0}E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{\sigma}^{2}\right] - \cdots \\ -p_{M^{2}-1}E\left[\|\bar{\boldsymbol{w}}_{i}\|_{FM^{2}-1\sigma}^{2}\right] \\ + \mu^{2}\sigma_{v}^{2}E\left[\|\overline{\mathbf{u}}_{i}\|_{FM^{2}-1\sigma}^{2}\right] \\ + \mu^{2}\beta^{2}\|\boldsymbol{w}^{\circ}\|_{FM^{2}-1\sigma}^{2} \\ + \mu^{2}\beta^{2}\|\boldsymbol{w}^{\circ}\|_{FM^{2}-1\sigma}^{2}\right] \\ + \mu^{2}\beta^{2}\|\boldsymbol{w}^{\circ}\|_{FM^{2}-1\sigma}^{2} \\ + \mu^{2}\beta^{2}\|\boldsymbol{w}^{\circ}\|_{FM^{2}-1\sigma}^{2} \\ + \mu^{2}\beta^{2}\|\boldsymbol{w}^{\circ}\|_{FM^{2}-1\sigma}^{2}\right] \end{cases}$$

In the above system of equations, f_k is a vector defined by $f_k = J \mathbf{L}_k \boldsymbol{w}^o$, where \mathbf{L}_k is matrix of size M such that vec $(\mathbf{L}_k) = F^k \boldsymbol{\sigma}$. The last expression in the above system is obtained from the previous one by means of the Cayley-Hamilton theorem, which enables us to express F^{M^2} as a linear combination of lower powers:

$$\boldsymbol{F}^{M^2} = -p_0 \boldsymbol{I} - p_1 \boldsymbol{F} - \dots - p_{M^2 - 1} \boldsymbol{F}^{M^2 - 1}$$

where the p_i 's are the coefficients of the characteristic polynomial of F, viz., $p(x) = \det(xI - F)$.

In summary, recursions (15)-(18) can be combined together into a single matrix recursion in state-space form:

$$\begin{bmatrix} \mathcal{W}_{i+1} \\ E[\bar{\boldsymbol{w}}_{i+1}] \end{bmatrix} = \begin{bmatrix} \mathbf{G}_1 & \mathbf{G}_2 \\ O & \boldsymbol{J} \end{bmatrix} \begin{bmatrix} \mathcal{W}_i \\ E[\bar{\boldsymbol{w}}_i] \end{bmatrix} + \begin{bmatrix} \mathcal{Y} \\ \mu\beta\boldsymbol{w}^o \end{bmatrix}$$
(19)

where the matrices $\{G_1, G_2\}$ are defined by

$$\mathbf{G_1} \triangleq \begin{bmatrix} 0 & 1 & \cdots & 0 \\ 0 & 0 & \cdots & 0 \\ \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & \cdots & 1 \\ -p_0 & -p_1 & \cdots & -p_{M^2-1} \end{bmatrix}$$
(20)

$$\mathbf{G}_{2} \stackrel{\Delta}{=} 2\mu\beta \begin{bmatrix} \boldsymbol{f}_{0}^{T} \\ \boldsymbol{f}_{1}^{T} \\ \vdots \\ \boldsymbol{f}_{M^{2}-1}^{T} \end{bmatrix}, \quad \mathcal{W}_{i} \stackrel{\Delta}{=} \begin{bmatrix} E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{F}^{2}\right] \\ E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{F\sigma}^{2}\right] \\ \vdots \\ E\left[\|\tilde{\boldsymbol{w}}_{i}\|_{FM^{2}-1\sigma}^{2}\right] \end{bmatrix}$$

and

$$\mathcal{Y} = \mu^{2} \begin{bmatrix} \sigma_{v}^{2} E \left[\| \overline{\mathbf{u}}_{i} \| _{\sigma}^{2} \right] + \beta^{2} \| \boldsymbol{w}^{o} \| _{\sigma}^{2} \\ \sigma_{v}^{2} E \left[\| \overline{\mathbf{u}}_{i} \| _{F\sigma}^{2} \right] + \beta^{2} \| \boldsymbol{w}^{o} \| _{F\sigma}^{2} \\ \vdots \\ \sigma_{v}^{2} E \left[\| \overline{\mathbf{u}}_{i} \| _{FM^{2}-1\sigma}^{2} \right] + \beta^{2} \| \boldsymbol{w}^{o} \| _{FM^{2}-1\sigma}^{2} \end{bmatrix}$$

The state recursion (19) characterizes the transient behavior of the leaky adaptive filters (1)-(2). It can now be used to study mean-square stability and mean-square error performance.

3.1. Stability

From (19), we see that stability is achieved if, and only if, both G_1 and J are stable matrices. However, since G_1 and F have the same eigenvalues, this condition corresponds to requiring that F and J be stable matrices. By inspecting the defining expressions (12) for J and (16) for F, we can show that

$$J ext{ is stable } \Leftrightarrow \ \mu < rac{2}{\lambda_{\max}\left(E\left[\mathcal{U}_{i}
ight]
ight)}$$
 (21)

$$F ext{ is stable } \Leftrightarrow \mu < rac{1}{\lambda_{\max} \left(\mathbf{A}^{-1} \mathbf{B}
ight)}$$
 (22)

where $\mathbf{A} = E[\mathcal{U}_i] \otimes \mathbf{I} + \mathbf{I} \otimes E[\mathcal{U}_i]$ and $\mathbf{B} = E[\mathcal{U}_i \otimes \mathcal{U}_i]$.

3.2. Steady-State Error

Steady-state performance can be obtained directly from recursion (17). So, assuming the filter is stable, we get $E\left[\|\tilde{w}_{i+1}\|_{\sigma}^2\right] = E\left[\|\tilde{w}_i\|_{\sigma}^2\right]$ as $i \to \infty$. Therefore, in the limit, relations (17) and (18) lead to

$$\begin{split} \lim_{i \to \infty} E\left[\|\tilde{\boldsymbol{w}}_i\|_{\sigma}^2 \right] &= \lim_{i \to \infty} E\left[\|\tilde{\boldsymbol{w}}_i\|_{F\sigma}^2 \right] + \mu^2 \sigma_v^2 E\left[\|\overline{\mathbf{u}}_i\|_{\sigma}^2 \right] \\ &+ \mu^2 \beta^2 \|\boldsymbol{w}^o\|_{\sigma}^2 + 2\mu\beta \boldsymbol{w}^{o^T} \boldsymbol{\Sigma} \boldsymbol{J} \lim_{i \to \infty} E\left[\tilde{\boldsymbol{w}}_i \right] \end{split}$$

and

$$\lim_{i\to\infty} E\left[\tilde{\boldsymbol{w}}_i\right] = \mu\beta\left(\boldsymbol{I}-\boldsymbol{J}\right)^{-1}\boldsymbol{w}^o$$

or, equivalently,

$$\lim_{i \to \infty} E\left[\|\tilde{\boldsymbol{w}}_i\|_{(I-F)\sigma}^2 \right] = \mu^2 \sigma_v^2 E\left[\|\tilde{\mathbf{u}}_i\|_{\sigma}^2 \right] + \mu^2 \beta^2 \|\boldsymbol{w}^o\|_{\sum (I+2J(I-J)^{-1})}^2$$
(23)

This expression allows us to evaluate the steadystate weight-error energy for any choice of a symmetric weight Σ . In particular, we can get the mean-square error by choosing $\Sigma = \mathbf{R}$, i.e., by choosing σ such that $(I - F)\sigma = \operatorname{vec}(\mathbf{R})$. This leads to the expression

$$\lim_{i \to \infty} E\left[e_a^2(i)\right] = \mu^2 \sigma_v^2 E\left[\|\overline{\mathbf{u}}_i\|_{(I-F)^{-1} \operatorname{vec}(R)}^2 \right] \\ + \mu^2 \beta^2 \|\boldsymbol{w}^o\|_{\Sigma\left(I+2J(I-J)^{-1}\right)}^2$$

where $\operatorname{vec}(\Sigma) = (I - F)^{-1}\operatorname{vec}(R)$. Similarly, the meansquare deviation is obtained by choosing σ (and hence Σ) such that $(I - F)\sigma = \operatorname{vec}(I)$.

4. TRACKING ANALYSIS

The results of the previous section can be specialized for non-leaky normalized filters by setting $\alpha = \beta = 0$. More importantly, the analysis can be used to infer (almost immediately) the tracking performance of normalized adaptive filters. In the tracking case, w^{o} is no more constant but undergoes random perturbations, say

$$\boldsymbol{w}_{i+1}^o = \boldsymbol{w}_i^o + \boldsymbol{q}_i$$

As in the leaky case, we still carry out the derivation in terms of the error quantities $e_a(i)$ and \tilde{w}_i as defined in (4), with w^o replaced by the now time varying w_i^o . To perform mean-square analysis in the tracking case, we rely on assumptions AN and AI, in addition to the following assumption:

AT The sequence of tracking errors $\{q_i\}$ is zero-mean and stationary, and is independent of the input \mathbf{u}_i and the additive noise v(i).

Now consider the adaptation equation (1) for $\alpha = 0$, rewritten in terms of $\overline{\mathbf{u}}_i$, $e_a(i)$, and $\tilde{\boldsymbol{w}}_i$:

$$\bar{\boldsymbol{w}}_{i+1} = \bar{\boldsymbol{w}}_i - \mu e(i) \overline{\mathbf{u}}_i^T + \boldsymbol{q}_i \tag{24}$$

Notice that this is the same as (5) for $\alpha = 0$, $\beta = 1/\mu$, and $w^o = q_i$. We can similarly argue that the mean-square behavior is also described by (17) for the same values of α and β and for¹

$$\|\boldsymbol{w}^{o}\|_{\sigma}^{2} = E\left[\|\boldsymbol{q}\|_{\sigma}^{2}\right], \text{ and } \boldsymbol{w}^{o} = E\left[\boldsymbol{q}_{i}\right] = 0$$

That is, we now have

$$E\left[\left\|\tilde{\boldsymbol{w}}_{i+1}\right\|_{\sigma}^{2}\right] = E\left[\left\|\tilde{\boldsymbol{w}}_{i}\right\|_{F\sigma}^{2}\right] + \mu^{2}\sigma_{v}^{2}E\left[\left\|\boldsymbol{u}_{i}\right\|_{\sigma}^{2}\right] + E\left[\left\|\boldsymbol{q}_{i}\right\|_{\sigma}^{2}\right]$$
(25)

Stability and steady-state behavior can now be deduced from (25). In particular, (mean-square) stability is guaranteed if, and only if, F is a stable matrix (see (22)) where now

$$\mathcal{U}_{i} = \alpha I + \mathbf{u}_{i}^{T} \overline{\mathbf{u}}_{i} \Big|_{\alpha=0} = \mathbf{u}_{i}^{T} \overline{\mathbf{u}}_{i}$$

Moreover, by an approach similar to that of the previous section, we can derive the following expression for the steady-state error

$$\lim_{i \to \infty} E\left[\left\| \tilde{\boldsymbol{w}}_i \right\|_{\sigma}^2 \right] = \mu^2 \sigma_v^2 E\left[\left\| \overline{\mathbf{u}}_i \right\|_{(I-F)^{-1}\sigma}^2 \right] + E\left[\left\| \boldsymbol{q}_i \right\|_{(I-F)^{-1}\sigma}^2 \right]$$

5. CONCLUSION

In this paper, we performed mean-square analysis of leaky normalized adaptive filters. We showed how the analysis can be further used to infer the tracking performance of normalized adaptive filters. Our study applies to a large class of data nonlinearities and does not impose Gaussian assumptions on the data.

6. REFERENCES

- K. Mayyas and T. Aboulnasr, "Leaky LMS algorithm: MSE analysis for Gaussian data," *IEEE Trans. on Signal Processing*, vol. 45, no. 4, pp. 927-934, Apr. 1997.
- [2] V. H. Nascimento and A. H. Sayed, "Unbiased and stable leakage-based adaptive filters," *IEEE Transactions* on Signal Processing, vol. 47, no. 12, pp. 3261–3276, Dec. 1999.
- [3] A. H. Sayed and M. Rupp, "A time-domain feedback analysis of adaptive algorithms via the small gain theorem," *Proc. SPIE*, vol. 2563, pp. 458-69, San Diego, CA, Jul. 1995.
- [4] M. Rupp and A. H. Sayed, "A time-domain feedback analysis of filtered-error adaptive gradient algorithms," *IEEE Transactions on Signal Processing*, vol. 44, no. 6, pp. 1428-1439, Jun. 1996.
- [5] N. R. Yousef and A. H. Sayed, "A unified approach to the steady-state and tracking analyses of adaptive filters," *IEEE Transactions on Signal Processing*, vol. 49, no. 2, pp. 314-324, February 2001. [See also Proc. 4th *IEEE-EURASIP Workshop on Nonlinear Signal and Image Processing*, vol. 2, pp. 699-703, Antalya, Turkey, June 1999.]
- [6] T. Y. Al-Naffouri and A. H. Sayed, "Transient analysis of adaptive filters," *Proc. ICASSP*, Salt Lake City, Utah, May 2001.
- [7] N. J. Bershad, "Analysis of the normalized LMS algorithm with Gaussian inputs," *IEEE Trans. Acoust. Speech Signal Process.*, vol. 34, pp. 793–806, 1986.
- [8] M. Rupp, "The behavior of LMS and NLMS algorithms in the presence of spherically invariant processes," *IEEE Transactions on Signal Processing*, vol. 41, no. 3, March 1993.

¹For completeness, we point out that the mean weight-error behavior can similarly be obtained from (18) with $\alpha = 0, \beta = 1/\mu$, and $w_o = E[q_i] = 0$.