## King Fahd University of Petroleum \& Minerals

Electrical Engineering Department
EE370: Communications Engineering I (071)

## Major Exam I

October 28, 2007
7:00 PM-8:30PM
Building 59-2001/2002

Name: $\qquad$ KEY $\qquad$
ID\# : $\qquad$

Please write your Serial \#

Section (3) Dr. Kousa, SMW 9:00-9:50am
Please circle your section
(2) Dr. Kousa, SMW 10:00-10:50am
(1) Dr. Muqaibel, UT 8:30-9:45am
(4) Dr. Muqaibel, UT 10:00-11:15am

| Question | Mark |
| :---: | :---: |
| 1 | $/ 10$ |
| 2 | $/ 12$ |
| 3 | $/ 12$ |
| 4 | $/ 11$ |
| Total | $/ 45$ |

## Instructions:

1. This is a closed-book/notes exam.
2. The duration of this exam is one and half hours.
3. Read the questions carefully. Plan which question to start with.
4. CLEARLY LABEL ALL SIGNIFICANT VALUES ON BOTH AXIES OF ANY SKETCH
5. Work on your own.
6. Mobile phones are not allowed in the exam room.
7. Tables of Fourier Transform pairs, FT properties and Trigonometric Identities are provided in the last sheet.

## Problem 1: ( 10 points)

Clearly write True (T) or False (F).

| a | The antenna dimension is inversely proportional to the transmitting frequency | $\mathbf{T}$ |
| :--- | :--- | :--- |
| $\mathbf{b}$ | $x^{2} \delta(x-1) \neq 1$ | T |
| $\mathbf{c}$ | For a distortionless system, the amplitude spectrum must be constant, and the phase <br> must be a linear function of frequency. | $\mathbf{T}$ |
| $\mathbf{d}$ | For a modulating signal that is real, the SSB modulated signal is complex. | F |
| $\mathbf{e}$ | Envelope detection is a type of coherent demodulation. | F |
| $\mathbf{f}$ | AM (DSB with carrier) provides saving in the bandwidth at the cost of higher <br> transmitted power. | F |
| $\mathbf{g}$ | The Hilbert transform of the Hilbert transform of a signal is the negative of the signal <br> itself. | $\mathbf{T}$ |
| $\mathbf{h}$ | To allow for non-coherent detection in AM (DSB with carrier) the modulation index <br> should be greater than 1 | F |
| $\mathbf{i}$ | The energy of a power signal is finite. | F |
| $\mathbf{j}$ | The bandwidth efficiency of Quadrature Amplitude Modulation (QAM) is the same as <br> that of DSBSC modulation | $\mathbf{F}$ |

Problem 2: ( 12 points)
Consider the function $g(t)$ shown to the right.
Let its FT be denoted by $G(\omega)$.


Express $g_{1}(t), g_{2}(t)$ and $g_{3}(t)$ in terms of $g(t)$ and find their spectra in terms of $G(\omega)$.




$$
\begin{aligned}
& g_{3}(t)=g(t)+g(-t-1)-g\left(-C_{a}(\omega) e^{-j \omega}\right. \\
& G_{3}(\omega)=G_{a}(-\omega) e^{5 \omega}-G(-\omega)-C_{a}(\omega)
\end{aligned}
$$

$$
\begin{aligned}
& g_{\text {Temp }}(t)=g(t-1) \\
& \xrightarrow[1]{\longrightarrow} \\
& g_{\operatorname{tin}_{2}}(t)=g\left(\frac{t}{2}-1\right) \\
& g_{1}(t)=1.5 g\left(\frac{t}{2}-1\right)=1.5 g\left(\frac{t-2}{2}\right) \\
& g_{2}(t)=g(t-1)+g(-t-1)=g(t-1)+g(-(t+1)) \\
& \text { other Correct } \\
& \text { solutions are } \\
& \text { also possible. } \\
& g_{3}(t)=\frac{d}{d t} g_{2}(t)=\frac{d}{d t}[g(t-1)+g(-t-1)] \\
& G_{\operatorname{tenp}}^{(\omega)}=G(\omega) e^{J \omega} \\
& G_{1}(\omega)=1.5(2) G(2 \omega) e^{-j \omega(+2)} \\
& G_{\tan _{2}}(\omega)=G(2 \omega) e^{J 2 \omega} \\
& G_{1}(\omega)=3 G(2 \omega) e^{-J \omega 2} \\
& G_{2}(\omega)=G(\omega) e^{-j \omega(+1)}+G(-\omega) e^{+j \omega} \\
& G_{3}(\omega)=J \omega G_{2}(\omega) \\
& =J \omega\left[G(\omega) e^{-j \omega}+G(-\omega) e^{J \omega}\right] \\
& \text { We may also write. } \\
& \frac{\text { We may also write }}{g_{3}(t)=g(t)+g(-t-i)}-g(-t)-g(t-1) \\
& \underbrace{}_{\substack{g(t)+g(-t-1) \\
l}} \\
& \begin{array}{l}
g(t-1) \\
-g(-t)
\end{array}
\end{aligned}
$$

Problem 3: Amplitude Modulation (12 points)
An AM (DSB with carrier) modulator operates with the message signal:

$$
m(t)=4 \cos (2 \pi 10 t)+6 \cos (2 \pi 30 t)
$$

The un-modulated carrier is given by $\cos (2 \pi 1000 t)$
Hint: the maximum and minimum values of $m(t)$ are $10,-10$, respectively
(a) If the modulation index is set to 0.5 ,
(b) (i) Write an expression for the modulated signal, $\varphi_{A M}(t)$.
(ii) Sketch the magnitude spectrum of the modulator output, $\varphi_{A M}(t)$, showing the values at all frequencies of interest.
(iii) Determine the modulator efficiency, $\eta$, defined by the ratio between the power of sidebands (message) and the total power.
(b) If envelope detection is going to be used for the detection of the above modulated signals, design proper value for the circuit time constant $R C$.

$$
\begin{equation*}
\mu=\frac{m_{p}}{A}=\frac{10}{A}=0.5 \Rightarrow A=20 \tag{0}
\end{equation*}
$$

(a)
(2)

$$
\begin{aligned}
\varphi_{A M}(t) & =[m(t)+A] \cos w_{c} t \\
& =[4 \cos (2 \pi 10 t)+6 \cos (2 \pi 30 t)+2] \cos (2 \pi 1000 t)
\end{aligned}
$$

$$
M(w)
$$

(iv)

$i i i)=$

$$
\eta=\frac{13 \times 100}{213} \%=6.1 \%
$$

b)

$$
\begin{aligned}
& \frac{1}{w_{c}} \ll R C \ll \frac{1}{2 \pi B}<\text { This is supposed to be } \\
& \frac{1}{2 \pi(1000)} \ll R C \ll \frac{1}{2 \pi(30)} \\
& 1.59 \times 10^{-4} \ll R C \ll .3 \times 10^{-3} \\
& \text { in batwean. } 1 \times 10^{-3} \mathrm{sec}
\end{aligned}
$$

Problem 4: (11 points)
2 (a) The impulse response of a Hilbert transform is $1 /(\pi t)$. Derive its transfer function $H(\omega)$.
(b) Consider the phase shift method of generating LSB-SSB waves.

4 (i) Sketch the block diagram of the system.
5 (ii) Let the input to the system be the message $\cos (2 \pi t)$, and let the carrier signal be $\cos (20 \pi t)$. Trace the signal throughout the system mathematically and graphically in the TIME DOMAIN. (That is, write the expression and sketch the signal after each stage).
sign $(t) \leftrightarrow \frac{2}{J w} \quad$ for $g(t) \leftrightarrow G(w), G(t) \leftrightarrow 2 \pi g(-\omega)$
so $j \frac{\operatorname{syn} t}{2 \pi} \leftarrow \frac{1}{\pi \omega}$

$$
\frac{1}{\pi t} \longleftrightarrow 2 \pi \frac{j}{2 \pi} \operatorname{sgn}(-\omega)=-\jmath \operatorname{sgn}(\omega)
$$

$\operatorname{sgn}(-\omega)=-\operatorname{sgn}(\omega)$ odd function
(a), application of duality property to pair 12 of Table 3.1 "shout tale of Fourier Transform Gilds. $\frac{1}{\pi t} \Leftrightarrow-J \operatorname{sgn}(\omega)$
(b)

(d) $\begin{aligned} & \sin 2 \pi t \sin 20 \pi t \\ & \left.=\frac{1}{2}[\cos (18 \pi t)-\cos (22 \pi t)]\right]\end{aligned}$
at a

$$
m(t)=\cos 2 \pi t
$$

$$
f_{m}=1 \mathrm{~Hz}
$$


(b)
$\cos \left(2 \pi t-\frac{\pi}{2}\right)=\sin (2 \pi t)$

(c)

$$
\left[\cos 2 \pi t \cos (20 \pi t)=\frac{1}{2}[\cos (18 \pi t)+\cos (22 t)]\right.
$$




| Short Table of Fourier Transforms |  |  |  | Trigonometric Identities |
| :---: | :---: | :---: | :---: | :---: |
|  | $g(t)$ | $G(\omega)$ |  |  |
| 1 | $e^{-a t} u(t)$ | $\frac{1}{a+j \omega}$ | $a>0$ |  |
| 2 | $e^{a t} u(-t)$ | $\frac{1}{a-j \omega}$ | $a>0$ | $\cos \mathrm{A} \cos \mathrm{B}=1 / 2[\cos (\mathrm{~A}+\mathrm{B})+\cos (\mathrm{A}-\mathrm{B})]$ |
| 3 | $e^{-a\|t\|}$ | $\frac{2 a}{a^{2}+\omega^{2}}$ | $a>0$ |  |
| 4 | $t e^{-a f_{u}} u(t)$ | $\frac{1}{(a+j \omega)^{2}}$ | $a>0$ | $\sin \mathrm{A} \sin \mathrm{B}=1 / 2[\cos (\mathrm{~A}-\mathrm{B})-\cos (\mathrm{A}+\mathrm{B})]$ |
| 5 | $t^{\prime \prime} e^{-a t} u(t)$ | $\frac{n!}{(a+j \omega)^{n+1}}$ | $a>0$ | $\sin A \cos B=1 / 2[\sin (A+B)+\sin (A-B)]$ |
| 6 | $\delta(t)$ | 1 |  |  |
| 7 | 1 | $2 \pi \delta(\omega)$ |  |  |
| 8 | $e^{j o n t}$ | $2 \pi \delta\left(\omega-\omega_{0}\right)$ |  |  |
| 9 | $\cos \omega_{0} t$ | $\pi\left[\delta\left(\omega-\omega_{0}\right)+\delta\right.$ |  |  |
|  | $\sin \omega_{0} t$ | $j \pi\left[\delta\left(\omega+\omega_{0}\right)-\right.$ |  |  |
|  |  | $\pi \delta(\omega)+\frac{1}{j \omega}$ |  |  |
|  | $\operatorname{sgn} t$ | $\frac{2}{j \omega}$ |  |  |
| 13 | $\cos \omega_{0} t u(t)$ | $\frac{\pi}{2}\left[\delta\left(\omega-\omega_{0}\right)+\right.$ |  |  |
| 14 | $\sin \omega_{0} t u(t)$ | $\frac{\pi}{2 j}\left[\delta\left(\omega-\omega_{0}\right)-\right.$ |  |  |
| 15 | $e^{-a t} \sin \omega_{0} t u(t)$ | $\frac{\omega_{0}}{(a+j \omega)^{2}+\omega_{0}^{2}}$ | $a>0$ |  |
|  | $e^{-a t} \cos \omega_{0} t u(t)$ | $\frac{a+j \omega}{(a+j \omega)^{2}+\omega_{0}^{2}}$ | $a>0$ |  |
|  | $\operatorname{rect}\left(\frac{t}{\tau}\right)$ | $\tau \operatorname{sinc}\left(\frac{\omega \tau}{2}\right)$ |  |  |
|  | $\frac{W}{\pi} \operatorname{sinc}(W t)$ | $\operatorname{rect}\left(\frac{\omega}{2 W}\right)$ |  |  |
| Fourier Transform Operations |  |  |  |  |
| Operation |  | $g(t)$ |  | $G(\omega)$ |
| Addition <br> Scalar multiplication Symmetry |  | $g_{1}(t)+g_{2}(t)$ |  | $G_{1}(\omega)+G_{2}(\omega)$ |
|  |  | $k g(t)$ |  | $k G(\omega)$ |
|  |  | $G(t)$ |  | $2 \pi g(-\omega)$ |
| Scaling |  | $g(a t)$ |  | $\frac{1}{\|a\|} G\left(\frac{\omega}{a}\right)$ |
| Time shift |  | $g\left(t-t_{0}\right)$ |  | $G(\omega) e^{-j \omega t_{0}}$ |
| Frequency shift |  | $g(t) e^{j \omega_{0} t}$ |  | $G\left(\omega-\omega_{0}\right)$ |
| Time convolution |  | $g_{1}(t) * g_{2}(t)$ |  | $G_{1}(\omega) G_{2}(\omega)$ |
| Frequency convolution |  | $g_{1}(t) g_{2}(t)$ |  | $\frac{1}{2 \pi} G_{1}(\omega) * G_{2}(\omega)$ |
| Time differentiation |  | $\underline{d^{n} g}$ |  | $(j \omega)^{n} G(\omega)$ |
| Time integration |  | $\int_{-\infty}^{t} g(x) d x$ |  | $\frac{G(\omega)}{j \omega}+\pi G(0) \delta(\omega)$ |

