

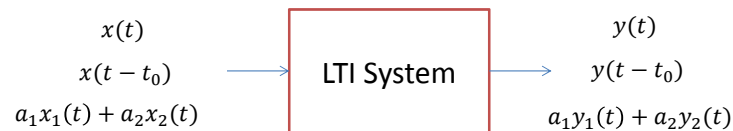
# Continuous-Time Linear Time-Invariant (LTI) Systems

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## Introduction

- **Why LTI?**
  - Many physical systems.
  - Easy to solve mathematically
  - Available information about analysis and design.
  - We can apply superposition



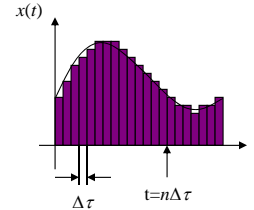
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## Impulse representation of Continuous Time Systems

- A signal can be represented as a sum of deltas (impulses).

$$x(t) \approx \sum_{n=-\infty}^{\infty} x(n\Delta\tau) \operatorname{rect}\left(\frac{t-n\Delta\tau}{\Delta\tau}\right)$$

$$x(t) = \lim_{\Delta\tau \rightarrow 0} \sum_{n=-\infty}^{\infty} x(n\Delta\tau) \operatorname{rect}\left(\frac{t-n\Delta\tau}{\Delta\tau}\right) = \int_{-\infty}^{\infty} x(\tau) \delta(t-\tau) d\tau$$



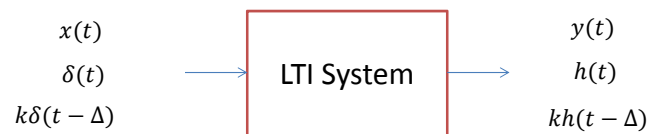
- To verify  $x(t) = \int_{-\infty}^{+\infty} x(\tau) \delta(t-\tau) d\tau$
- $x(t) = \int_{-\infty}^{+\infty} x(\tau) \delta(-(\tau-t)) d\tau$
- $x(t) = \int_{-\infty}^{+\infty} x(\tau) \delta(\tau-t) d\tau$

$$\delta(t) = \delta(-t)$$

The well know sifting (sampling) property

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## The Impulse Response



- Impulse response** of a system is response of the system to an input that is a unit impulse (i.e., a Dirac delta function in continuous time)
- Therefore, we know how to calculate the system output for any input,  $x(t)$

$$x(t) = \int_{-\infty}^{\infty} x(\tau) \delta(t-\tau) d\tau \Rightarrow \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau = y(t)$$

- This operation is called **convolution**

$$y(t) = \int_{-\infty}^{+\infty} x(\tau) h(t-\tau) d\tau$$

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## Convolution Concept

- $x(t) = u(t)$
- $h(t) = e^{-t}u(t)$
- What is  $y(t)$ ?
- $x(t) \approx \sum_{k=0}^{\infty} \delta(t - \Delta k)$
- $y(t) \approx \sum_{k=0}^{\infty} e^{-(t-\Delta k)} u(t - \Delta k)$
- If we take the limit as  $\Delta \rightarrow 0$  we get  $y(t)$

```
% Dr. Ali Hussein Mqaibel
clear all
close all
clc
t_step=0.1;
t=0:t_step:10;
h=exp(-t).*Heaviside(t);
subplot(3,1,1)
plot(t,h)
xlabel('time, s')
ylabel('The impulse response')
axis([0 10 0 2])
y=h;
x=[1 zeros(1,length(h)-1)].*Heaviside(t);

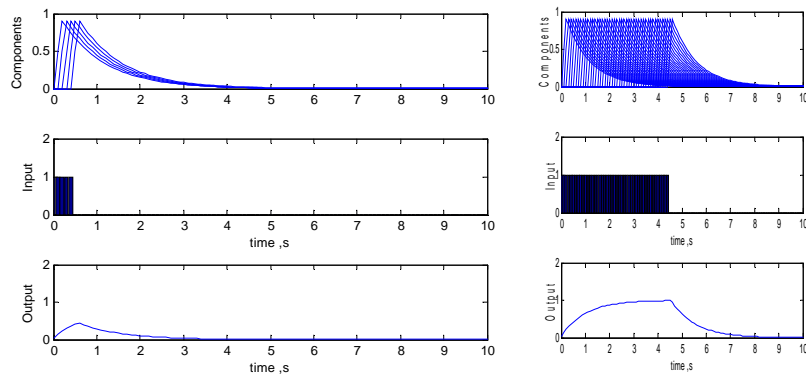
for i=1:length(t);
    subplot(3,1,1)

    plot(t,[zeros(1,i) h(1:end-i)])
    y=y+[zeros(1,i) h(1:end-i)];
    x(1,i)=1;
    hold on
    ylabel('Components')
    subplot(3,1,2)
    bar(t,x)
    xlabel('time, s')
    ylabel('Input')
    axis([0 10 0 2])

    subplot(3,1,3)
    plot(t,y*t_step)
    xlabel('time, s')
    ylabel('Output')
    axis([0 10 0 2])
    pause
end
```

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## Convolution for Continuous Time LTI Systems



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## Some Convolution Properties

- $y(t) = \int_{-\infty}^{+\infty} x(\tau)h(t - \tau)d\tau = x(t) * h(t)$
- $y(t) = \int_{-\infty}^{+\infty} x(t - \tau)h(\tau)d\tau = h(t) * x(t)$
- **What is the output if the input is impulse ?**
  - $y(t) = \delta(t) * h(t) = h(t)$
- For LTI system
  - $y(t - t_0) = \delta(t - t_0) * h(t) = h(t - t_0)$
- Recall some delta properties (no convolution)
  - $\delta(t - t_0)g(t) = g(t_0)\delta(t - t_0)$
  - $g(t - t_0)\delta(t) = g(-t_0)\delta(t)$
- $h(t)$  contains a complete input-output description .i.e. if the impulse response is known , the system response to any input can be found

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## Example: Impulse Response of an Integrator

- $y(t) = \int_{-\infty}^t x(\tau)d\tau$
- **What is  $h(t)$ ?**
  - $h(t) = y(t)|_{x(t)=\delta(t)} = \int_{-\infty}^t \delta(\tau)d\tau = u(t)$
- **What is the output if the input is ramp? "ramp response"**
  - $y(t) = x(t) * h(t) = tu(t) * u(t) = \int_{-\infty}^{+\infty} \tau u(\tau)u(t - \tau)d\tau$
  - $y(t) = \begin{cases} \int_0^{+\infty} \tau u(t - \tau)d\tau & t > 0 \\ 0 & t < 0 \end{cases} = \begin{cases} \int_0^t \tau d\tau & t > 0 \\ 0 & t < 0 \end{cases} = \begin{cases} \frac{t^2}{2} & t > 0 \\ 0 & t < 0 \end{cases} = \frac{t^2}{2} u(t)$
  - **Verify from the system equation**
  - $y(t) = \int_{-\infty}^t x(\tau)d\tau = \int_{-\infty}^t \tau u(\tau)d\tau$
  - $y(t) = \begin{cases} \int_{-\infty}^t \tau d\tau & t > 0 \\ 0 & t < 0 \end{cases} = \frac{t^2}{2} u(t)$
- Common mistake to forget  $u(t)$
- Before you start graphical convolution visit: (The Joy of Convolution)
- <http://www.jhu.edu/~signals/convolve/index.html>

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## Graphical Convolution Methods

- From the convolution integral, convolution is equivalent to

$$f_1(t) * f_2(t) \equiv \int_{-\infty}^{\infty} f_1(\tau) f_2(t-\tau) d\tau$$

- Rotating one of the functions about the y axis
- Shifting it by  $t$
- Multiplying this flipped, shifted function with the other function
- Calculating the area under this product
- Assigning this value to  $f_1(t) * f_2(t)$  at  $t$

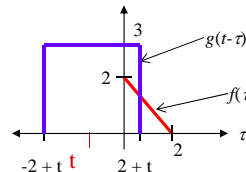
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## Graphical Convolution Example

- **Convolve the following two functions:**



- Replace  $t$  with  $\tau$  in  $f(t)$  and  $g(t)$
- Choose to flip and slide  $g(\tau)$  since it is simpler and symmetric
- Functions overlap like this:



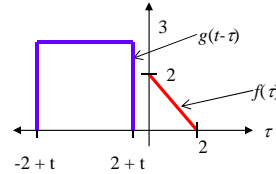
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## Graphical Convolution Example

- Convolution can be divided into 5 parts

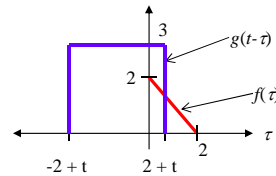
### I. $t < -2$

- Two functions do not overlap
- Area under the product of the functions is zero



### II. $-2 \leq t < 0$

- Part of  $g(t)$  overlaps part of  $f(t)$
- Area under the product of the functions is



$$\int_0^{2+t} 3(-\tau+2)d\tau = 3 \left( -\frac{\tau^2}{2} + 2\tau \right) \Big|_0^{2+t} = -\frac{3(2+t)^2}{2} + 6(2+t) = -\frac{3t^2}{2} + 6$$

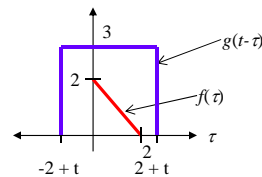
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## Graphical Convolution Example

### III. $0 \leq t < 2$

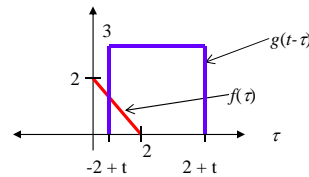
- Here,  $g(t)$  completely overlaps  $f(t)$
- Area under the product is just

$$\int_0^2 3(-\tau+2)d\tau = 3 \left( -\frac{\tau^2}{2} + 2\tau \right) \Big|_0^2 = 6$$



### IV. $2 \leq t < 4$

- Part of  $g(t)$  and  $f(t)$  overlap
- Calculated similarly to  $-2 \leq t < 0$



### V. $t \geq 4$

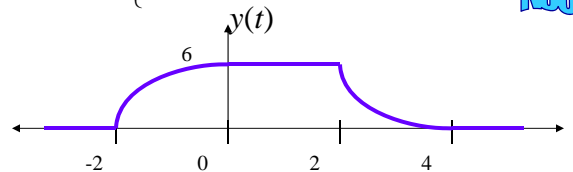
- $g(t)$  and  $f(t)$  do not overlap
- Area under their product is zero

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## Graphical Convolution Example

- Result of convolution (5 intervals of interest):

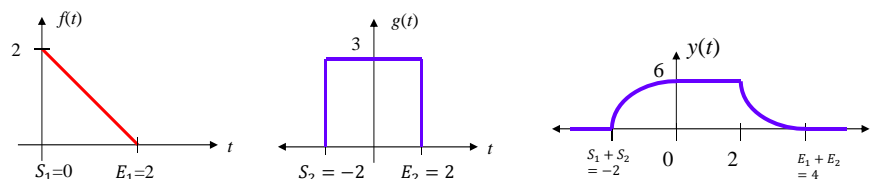
$$y(t) = f(t) * g(t) = \begin{cases} 0 & \text{for } t < -2 & \text{NoOverlap} \\ -\frac{3}{2}t^2 + 6 & \text{for } -2 \leq t < 0 & \text{PartialOverlap} \\ 6 & \text{for } 0 \leq t < 2 & \text{CompleteOverlap} \\ \frac{3}{2}t^2 - 12t + 24 & \text{for } 2 \leq t < 4 & \text{PartialOverlap} \\ 0 & \text{for } t \geq 4 & \text{NoOverlap} \end{cases}$$



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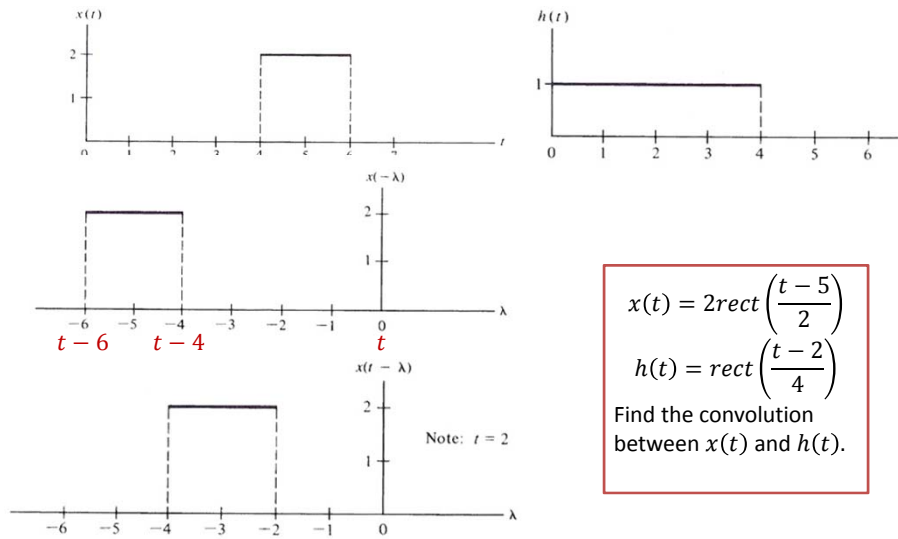
## How to Check the Convolution

- Start and end points: Start point equal to the sum of the starting points of the two signals and the end point equals to the sum of the end points.
- Area under the curve=product of the individual areas.



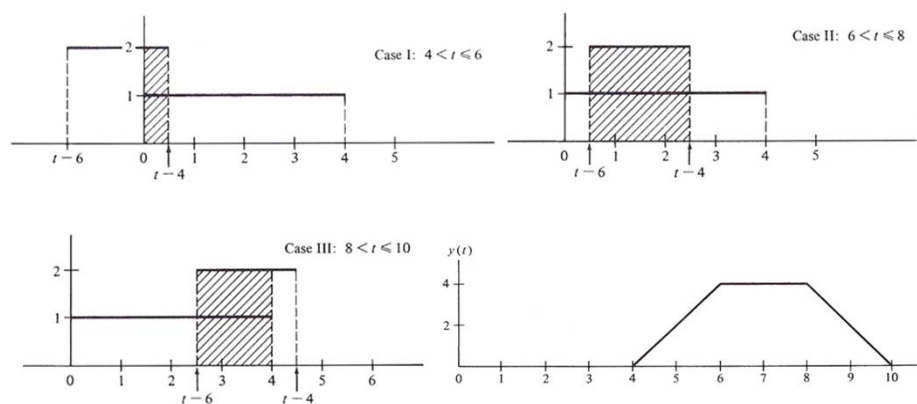
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## Example 2: Graphical Convolution



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## Continue Example 2:

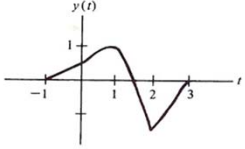


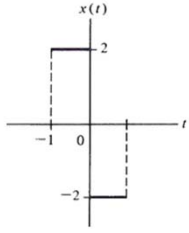
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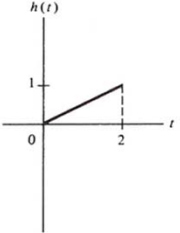


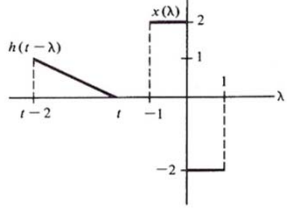
If we are interested in the output only at a specific time, we do not have to do full convolution

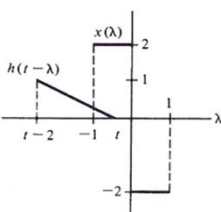
## Example 3:

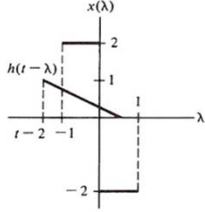


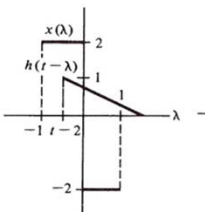


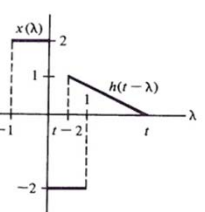












See additional Examples in the book .. You must write by your hand!

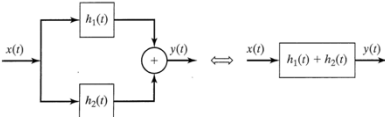
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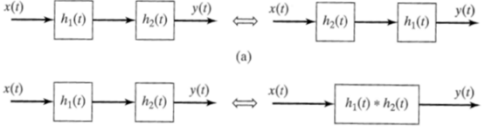
## Properties of Convolution

- Commutative property
  - $x(t) * h(t) = h(t) * x(t)$
- Associative Property
  - $[x(t) * h_1(t)] * h_2(t) = x(t) * [h_1(t) * h_2(t)]$
- Distributive property
  - $x(t) * [h_1(t) + h_2(t)] = x(t) * h_1(t) + x(t) * h_2(t)$

$x(t) \rightarrow \boxed{h(t)} \rightarrow y(t)$

$h(t) \rightarrow \boxed{x(t)} \rightarrow y(t)$

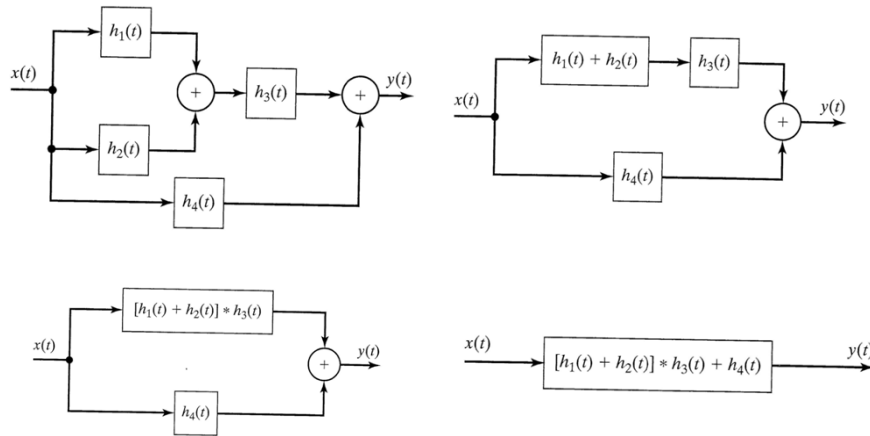




(a)

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## Example: Impulse Response of Interconnected System



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## Properties of Continuous-time LTI Systems

- The input-output characteristics of a continuous-time LTI system are completely described by its impulse response  $h(t)$
- $y(t) = \int_{-\infty}^{+\infty} x(\tau)h(t - \tau)d\tau$
- **Memory:**
  - In general LTI are dynamic because of integration over time.
  - The only way to get the integration out is if
  - $h(t) = k\delta(t) \Rightarrow y(t) = kx(t)$  ideal amplifier
  - An LTI system is memoryless if and only if  $h(t) = k\delta(t)$
- **Invertability**
  - $x(t) * h(t) * h_i(t) = x(t)$
  - $x(t) * \delta(t) = x(t)$
  - $h(t) * h_i(t) = \delta(t)$
  - An LTI system is invertible only if we can find  $h_i(t)$  such that  $h(t) * h_i(t) = \delta(t)$
  - No procedure is given to find  $h_i(t)$

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## Continue.. Properties of Continuous-time LTI Systems

- **Causality**

➤ A signal that is zero for  $t < 0$  is called a **causal signal**

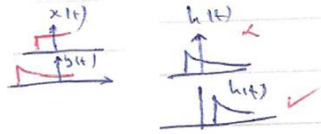
➤ Because  $\delta(t)$  occurs at  $t = 0 \Rightarrow h(t) = 0, t < 0$ .

➤  $y(t) = \int_{-\infty}^{+\infty} x(t - \tau)h(\tau)d\tau = \int_0^{+\infty} x(t - \tau)h(\tau)d\tau$

➤ Alternatively

➤  $h(t - \tau) = 0$  for  $t - \tau \Rightarrow \tau > t$

➤  $y(t) = \int_{-\infty}^{+\infty} x(\tau)h(t - \tau)d\tau = \int_{-\infty}^t x(\tau)h(t - \tau)d\tau$



Only for causal LTI

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## Continue..2.. Properties of Continuous-time LTI Systems

- **Stability**

• Since  $|\int_{-\infty}^{+\infty} x_1(t)x_2(t)dt| \leq \int_{-\infty}^{+\infty} |x_1(t)x_2(t)|dt$

• For a bounded input  $|x(t)| < M$  for all  $t$ , the output is

•  $|y(t)| = |\int_{-\infty}^{+\infty} x(t - \tau)h(\tau)d\tau| \leq \int_{-\infty}^{+\infty} |x(t - \tau)h(\tau)|d\tau$

•  $= \int_{-\infty}^{+\infty} |x(t - \tau)||h(\tau)|d\tau$

•  $\leq \int_{-\infty}^{+\infty} M|h(\tau)|d\tau$

•  $= M \int_{-\infty}^{+\infty} |h(\tau)|d\tau$

• i.e  $h(t)$  is absolutely integrable

•  $\int_{-\infty}^{+\infty} |h(t)|dt < \infty$

BIBO Stability test

$$\int_{-\infty}^{+\infty} |h(t)|dt < \infty$$

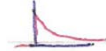
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## Stability Examples

- Example: Is the following system BIBO Stable?

- $h(t) = e^{-3t}u(t)$

- $\int_{-\infty}^{+\infty} |h(t)| dt = \int_0^{\infty} e^{-3t} dt = \frac{1}{3} < \infty$



- $h(t) = e^{+3t}u(t)$ .... Unstable

- $h(t) = u(t)$

- Integrator  $y(t) = \int_{-\infty}^t x(\tau) d\tau$

- $\int_0^{+\infty} |h(t)| dt = \int_0^{+\infty} dt = \rightarrow \infty$

- Unstable

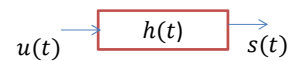
- Can be done by inspection of the graph of  $h(t)$

BIBO Stability test

$$\int_{-\infty}^{+\infty} |h(t)| dt < \infty$$

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## Unit Step Response



- If the input is  $u(t)$ , the output is noted as  $s(t)$  (some books they use  $a(t)$ )

- Relation between  $s(t)$  and  $h(t)$

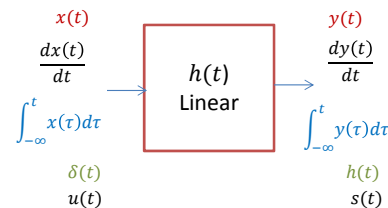
- $s(t) = \int_{-\infty}^{+\infty} u(\tau)h(t-\tau)d\tau = \int_0^{\infty} h(t-\tau)d\tau$

- $s(t) = \int_{-\infty}^{+\infty} h(\tau)u(t-\tau)d\tau = \int_{-\infty}^t h(t-\tau)d\tau = \int_0^t h(\tau)d\tau$

- $h(t) = \frac{ds(t)}{dt}$

- Note do not confuse  $\delta(t)$  with  $s(t)$  in writing !

- The unit step response completely describes the input-output characteristics of an LTI system.



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## Example: Step Response from the Impulse Response

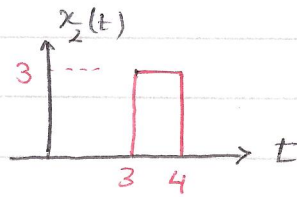
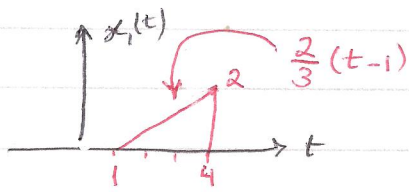
- $h(t) = e^{-3t}u(t)$ , find the step response.
- $s(t) = \int_0^t h(\tau)d\tau$
- $= \int_0^t e^{-3\tau}d\tau$
- $= \frac{e^{-3t}}{-3} \Big|_0^t = \frac{1}{3}(1 - e^{-3t})u(t)$
- Verify:
  - $h(t) = \frac{ds(t)}{dt} = e^{-3t}u(t) + \frac{1}{3}(1 - e^{-3t})\delta(t) = e^{-3t}u(t)$

Sifting property 0

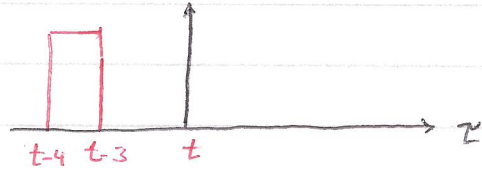
Practice more convolution examples. Consider both Graphical and analytical (See old exams and quizzes)

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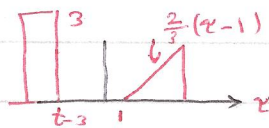
# Graphical Convolution Example:



$$y(t) = x_1(t) * x_2(t)$$



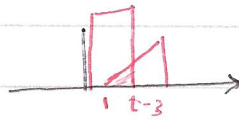
Case I no overlap



$$t-3 \leq 1 \Rightarrow t \leq 4$$

$$y(t) = 0$$

Case II partial overlap



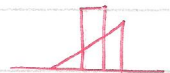
$$\int_1^{t-3} \frac{2}{3}(t-1)(3) d\tau = 2 \left[ \frac{\tau^2}{2} - \tau \right]_1^{t-3} = 2 \left[ \frac{(t-3)^2}{2} - (t-3) + \left( \frac{1}{2} - 1 \right) \right]$$

$$= 2 \left[ \frac{t^2 - 6t + 9}{2} - (t-3) + \frac{1}{2} \right] = t^2 - 8t + 16$$

$$5 \geq t \geq 4$$

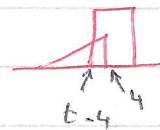
$$t-4 \leq 1 \Rightarrow t \leq 5$$

Case III Full overlap



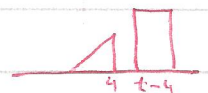
$$\int_{t-4}^{t-3} \frac{2}{3}(t-1)(3) d\tau = 2 \left[ \frac{\tau^2}{2} - \tau \right]_{t-4}^{t-3} = 2t - 9$$

Case IV  $t-3 \geq 4 \Rightarrow t \geq 7$  partial

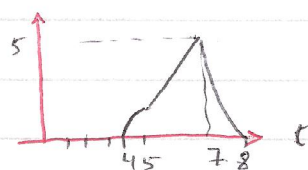


$$\int_{t-4}^4 \dots = -t^2 + 10t - 16$$

Case V  $t-4 > 4 \Rightarrow t > 8$  no overlap  $y(t) = 0$



$$y(t) = \begin{cases} 0 & t \leq 4 \\ t^2 - 8t + 16 & 5 \geq t > 4 \\ 2t - 9 & 7 \geq t > 5 \\ -t^2 + 10t - 16 & 8 \geq t > 7 \\ 0 & t \geq 8 \end{cases}$$



check start = 1 + 3 = 4 ✓

end = 4 + 4 = 8 ✓

Area,  $3 \times 4 = 12$  sk.