

**KING FAHD UNIVERSITY OF PETROLEUM AND MINERALS****Electrical Engineering Department****EE-205 Electric Circuits II****Spring 2009/2010(092) Second Major Exam****Duration : 90 min.****Dr. E. Hassan,****Dr. A. Muqaibel,****Dr. S. Al-Ghadban,****Dr. H. Masoudi (Coordinator)****Name :****KEY****ID # 000****Section # 0**

<b>Question</b>	<b>Grade</b>
<b>1 (10 points)</b>	<b>10</b>
<b>2 (10 points)</b>	<b>10</b>
<b>3 (10 points)</b>	<b>10</b>
<b>Total (30 points)</b>	<b>30</b>

**Notes :**

- 1) Read the question very carefully.**
- 2) Use a sketch to help you understand the question.**
- 3) Write neatly.**

Q1(a) For the circuit shown, write the matrix state equation

KVL for L (FL)

$$\frac{1}{5} \frac{di_L}{dt} + v_C - v = 0 \quad (1)$$

KCL for C (FCS)

$$\frac{1}{5} \frac{dv_C}{dt} + 2i - i_L = 0 \quad (2)$$

KVL for  $2\Omega$  (FL)

$$2i + v - v_s = 0 \quad (3) \Rightarrow v = v_s - 2i$$

State variables are  $v_C$  and  $i_L$

KCL for  $1\Omega$  (FCS)

$$v + i_L - i - 2i = 0 \quad (4)$$

(3) in (4)

$$v_s - 2i + i_L - 3i = 0$$

$$-5i = -v_s - i_L$$

$$i = \frac{v_s + i_L}{5} \quad (a)$$

(a) in (3)

$$v = v_s - 2\left(\frac{v_s + i_L}{5}\right)$$

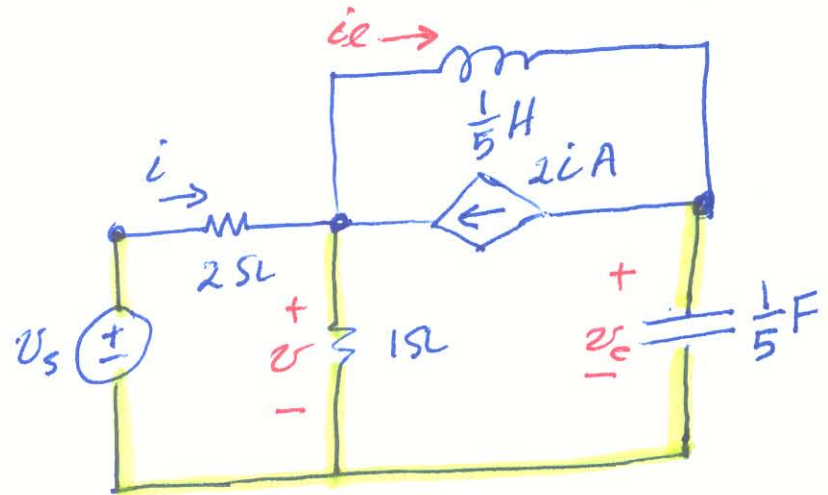
$$v = v_s - \frac{2v_s}{5} - \frac{2i_L}{5}$$

$$v = \frac{3v_s}{5} - \frac{2i_L}{5} \quad (b)$$

(a) and (b) in (1) and (2)

$$\frac{di_L}{dt} = -2i_L - 5v_C + 3v_s \quad (I)$$

$$\frac{dv_C}{dt} = 3i_L - 2v_s \quad (II)$$

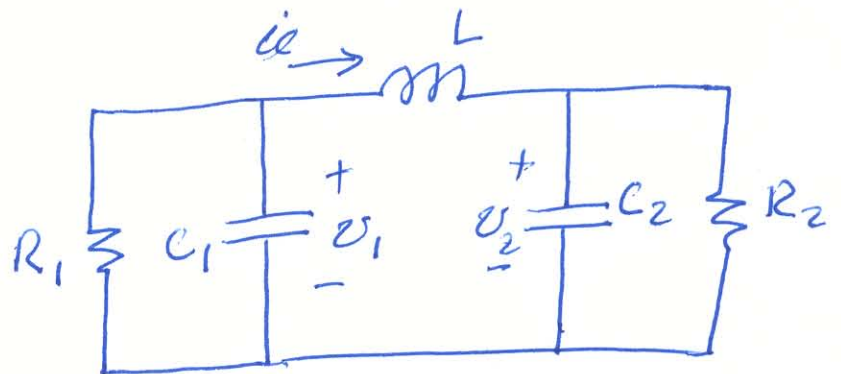


$$\frac{d}{dt} \begin{bmatrix} i_L \\ v_C \end{bmatrix} = \begin{bmatrix} -2 & -5 \\ 3 & 0 \end{bmatrix} \begin{bmatrix} i_L \\ v_C \end{bmatrix} + \begin{bmatrix} 3 \\ -2 \end{bmatrix} v_s$$

Q1 (b) For the circuit shown, the state equations of the circuit is given below. Given that  $R_1 = R_2 = 2\Omega$ ,  $C_1 = \frac{1}{2}F$  and  $C_2 = \frac{1}{2}F$  and  $L = 1H$ , assume that  $i_L(0) = 1A$ ,  $v_1(0) = 2V$  and  $v_2 = 0V$ .

i) Write the state equations in their numerical form using the numerical technique - (Euler's Method)

ii) Find the approximate values of  $i_L$ ,  $v_1$  and  $v_2$  at  $t = 0.1s$ ,  $t = 0.2s$  and  $t = 0.3s$



$$\frac{d}{dt} \begin{bmatrix} i_L \\ v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} 0 & \frac{1}{L} & -\frac{1}{L} \\ -\frac{1}{C_1} & \frac{1}{R_1 C_1} & 0 \\ \frac{1}{C_2} & 0 & \frac{1}{R_2 C_2} \end{bmatrix} \begin{bmatrix} i_L \\ v_1 \\ v_2 \end{bmatrix}$$

for the given values

$$\frac{d}{dt} \begin{bmatrix} i_L \\ v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 & -1 \\ -2 & -1 & 0 \\ 2 & 0 & 1 \end{bmatrix} \begin{bmatrix} i_L \\ v_1 \\ v_2 \end{bmatrix}$$

k	t	$i_L(t)$	$v_1(t)$	$v_2(t)$		
0	0.1					
1	0.2					
2	0.3					

Q1 (b) sol.

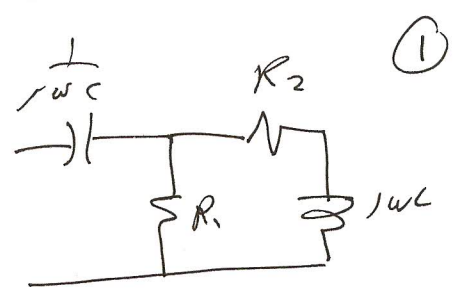
$$\begin{aligned}
 i) \quad \dot{v}_2([k+1]\Delta t) &\approx \dot{v}_2(k\Delta t) + [v_1(k\Delta t) - v_2(k\Delta t)] \Delta t \\
 v_1([k+1]\Delta t) &\approx v_1(k\Delta t) + [-2\dot{v}_2(k\Delta t) - v_1(k\Delta t)] \Delta t \\
 v_2([k+1]\Delta t) &\approx v_2(k\Delta t) + [2\dot{v}_2(k\Delta t) + v_2(k\Delta t)] \Delta t
 \end{aligned}$$

ii)

k	t(s)	$\dot{v}_2(t)$	$v_1(t)$	$v_2(t)$
0	0.1	1.2	1.6	0.2
1	0.2	1.34	1.2	0.46
2	0.3	1.414	0.812	0.774

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2	0.3	1.414	0.812	0.774

Find ~~(H)~~  $\omega_r$  for the circuit shown



$$Z_{in} = \frac{(R_2 + j\omega L)R_1}{R_1 + R_2 + j\omega L} + \frac{1}{j\omega c}$$

$$= \frac{R_1 (R_2 + j\omega L)(R_1 + R_2 - j\omega L)}{(R_1 + R_2)^2 + \omega^2 L^2} + \frac{1}{j\omega c}$$

$$= \frac{R_1}{(R_1 + R_2)^2 + \omega^2 L^2} [j\omega L(R_1 + R_2) - j\omega L R_2] + \frac{1}{j\omega c}$$

$$= \frac{j\omega L R_1^2}{(R_1 + R_2)^2 + \omega^2 L^2} = + \frac{j}{\omega c}$$

$$\therefore \omega^2 L C R_1^2 = \omega^2 L^2 + (R_1 + R_2)^2$$

$$\omega^2 (L C R_1^2 - L^2) = (R_1 + R_2)^2$$

$$\omega_r = \frac{R_1 + R_2}{\sqrt{L C R_1^2 - L^2}} = \frac{1}{\sqrt{L C}} \cdot \left( \frac{R_1 + R_2}{\sqrt{R_1^2 - L/C}} \right)$$

$$\text{if } R_1 \rightarrow \infty \rightarrow \omega_r \approx \frac{1}{\sqrt{L C}}$$

$$\omega_r = 3 \text{ rad/sec}$$

apply a test current source

$$Q = \frac{2\pi [W_L(t) + W_C(t)]_{\max}}{P_{RT}} \quad (0.5)$$

$$T = \frac{1}{f} = \frac{2\pi}{\omega} = \frac{2}{3}\pi \quad (0.5)$$

By CDR

$$\vec{I}_1 = 1 \angle 0^\circ \frac{10 - \frac{80}{3}j}{20 - \frac{80}{3}j} = 0.82 - j0.24 = 0.8544 \angle -16.314^\circ$$

$$\vec{I}_2 = 1 \angle 0^\circ \frac{10}{20 - \frac{80}{3}j} \quad (1)$$

$$\alpha \vec{I}_2 = 1 \angle 0^\circ - \vec{I}_1$$

$$\vec{I}_2 = 0.18 + j0.2 = 0.3 \angle 53.13^\circ$$

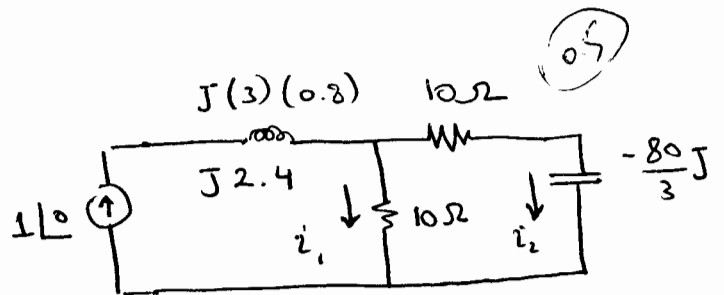
$$\vec{V}_C = \vec{I}_2 \frac{-80}{3}j = 8 \angle -36.87^\circ \quad (1)$$

$$W_C(t) = \frac{1}{2} C V_C^2(t)$$

$$= \frac{1}{2} \frac{1}{80} V_C^2(t)$$

$$= \frac{1}{160} (8)^2 \cos^2(3t - 36.87^\circ)$$

$$= 0.4 \cos^2(3t - 36.87^\circ) \quad (0.5)$$



$$i_s(t) = \cos 3t$$

$$W_L(t) = \frac{1}{2} L i_{\text{source}}^2 \quad (0.5)$$

$$= \frac{1}{2} (0.8) i_{\text{source}}^2(t)$$

$$= 0.4 \cos^2 3t$$

$$P_{R_1} = \frac{1}{2} (10) |I_1|^2 = \frac{1}{2} R_1 |I_1|^2 = 3.65$$

$$P_{R_2} = \frac{1}{2} (10) |I_2|^2 = 0.45 \quad (0.5)$$

$$Q = \frac{2\pi [0.4 \cos^2 3t + W_C(t)]_{\max}}{\frac{2}{3}\pi (3.65 + 0.45)}$$

$$= 0.7317 [W_L(t) + W_C(t)]_{\max}$$

$$= (0.7317)(0.4) [\cos^2 3t + \cos^2(3t - 36.87^\circ)]$$

$$= 0.2927 [1.8]$$

$$= 0.53 \quad (0.5)$$

How to find the maximum?

$$\cos^2 3t + \cos^2(3t - 36.87^\circ)$$

$$\frac{1}{2} [1 + \cos 6t] + \frac{1}{2} [1 + \cos(6t + 73.74^\circ)] \quad (1)$$

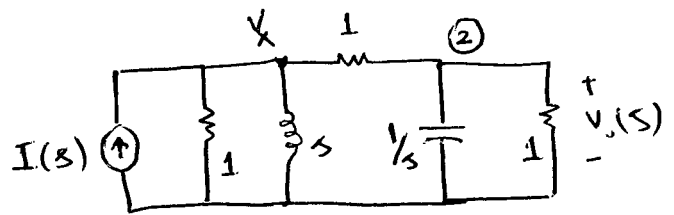
$$1 + \frac{\cos 6t + \cos(6t + 73.74^\circ)}{2}$$

$$\cos(\alpha + \beta) = \cos \alpha \cos \beta + \sin \alpha \sin \beta = 0.28 \cos 6t + 0.96 \sin 6t$$

$$1 + \frac{1.28 \cos 6t + 0.96 \sin 6t}{2}$$

$$\max 1 + \frac{1}{2} \sqrt{(1.28)^2 + (0.96)^2} = 1.8$$

assign the voltage  $V_x$  as shown



By KCL @ node  $V_x$

$$I = \frac{V_x}{1} + \frac{V_x}{s} + \frac{V_x - V_o}{1}$$

$$I + V_o = V_x + \frac{V_x}{s} + V_x = (2 + \frac{1}{s}) V_x = (\frac{2s+1}{s}) V_x$$

$$V_x = (I + V_o) \left( \frac{s}{2s+1} \right) \quad \text{--- ①}$$

By KCL @ node ②

$$\frac{V_x - V_o}{1} = \frac{V_o}{1/s} + \frac{V_o}{1}$$

$$V_x - V_o = sV_o + V_o \Rightarrow V_x = (s+2)V_o \quad \text{--- ②}$$

By Equating ① to ② to eliminate  $V_x$

$$(s+2)V_o = I \frac{s}{2s+1} + V_o \frac{s}{2s+1}$$

$$\left( s+2 - \frac{s}{2s+1} \right) V_o = I \frac{s}{2s+1} \quad , \text{ multiply both sides by } (2s+1)$$

$$\left[ (2s+1)(s+2) - s \right] V_o = s I$$

$$H(s) = \frac{V}{I} = \frac{s}{(2s+1)(s+2) - s} = \frac{s}{2s^2 + 3s + 4s + 2 - s} = \boxed{\frac{s}{2s^2 + 4s + 2}}$$

I) Given the following Transfer function

under-damped

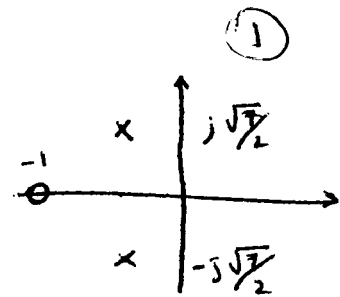
$$H(s) = \frac{V_2(s)}{V_1(s)} = \frac{2(s+1)}{s^2 + s + 2} = \frac{2(s+1)}{s^2 + s + 2}$$

a) Sketch the pole zero plot

zeros.  $s = -1$

poles

0.5



0.5

$$\frac{1^2 \pm \sqrt{1 - 8}}{2} = -\frac{1}{2} \pm j \frac{\sqrt{7}}{2}$$

The response is underdamped because the poles are two complex conjugate

b) if the input signal is  $v_1(t) = 10 e^{-t} \cos(t + \frac{\pi}{2})$

find the output voltage

1

$$s = -1 + j \quad \vec{V}_1 = 10 \angle \frac{\pi}{2} = 10 \angle 90^\circ$$

$$\vec{V}_2 = H(s) \Big|_{s = -1 + j} 10 \angle 90^\circ = (-1 + j) 10 \angle 90^\circ = \sqrt{2} \angle 135^\circ 10 \angle 90^\circ$$

1

$$V_2(t) = 14.14 e^{-t} \cos(t + 225^\circ)$$

$$= 14.14 \angle 225^\circ \text{ or } \angle -135^\circ$$

c) if the input is now  $v_1(t) = \sqrt{19} e^{-\frac{1}{2}t} \cos(\frac{\sqrt{7}}{2}t + 109^\circ)$

what would be the output voltage.

Justify your answer.

1

$s = -\frac{1}{2} + j \frac{\sqrt{7}}{2}$  this the location where

the transfer function has a pole

the output is expected to be  $\infty$  (theoretically).