

The previous example is an example of Linear Block codes. The parity bits are linear combination of the message. Therefore, we can represent the encoder by a linear system described by Matrices.

4.2.2 Representing Linear Codes in a Vector Space.

The 2-D code in Example 4.2.2 can be described using vector space.

codeword $\rightarrow \bar{C} = \bar{m}$

1	1	1	1	0	0	0	0	0	0	0	0	0	0	0	0
1	0	1	0	1	1	0	0	0	0	0	0	0	0	0	0
1	0	1	0	0	0	1	1	0	0	0	0	0	0	0	0
1	0	0	0	0	0	0	0	1	1	0	0	0	0	0	0
1	0	0	0	1	0	0	0	1	0	1	0	0	0	0	0
1	1	0	0	0	0	0	0	0	0	0	0	0	0	0	1
1	0	0	0	1	0	0	0	0	0	0	0	0	0	0	1
1	0	0	0	0	0	1	0	0	0	0	0	0	0	0	1

...

0	0	0
0	0	0
0	0	0
0	0	0
0	0	0
0	0	0
0	0	0
1	0	0
0	1	0
0	0	1

9x16

$$\bar{C} = \bar{m} G$$

\bar{C} : in 1x16 code word

\bar{m} : in 1x9 message bits

G : 9x16 Generator Matrix

~~At~~ At the receiver, we need to find the syndrome bits.

syndrome vector $\bar{s} = [s_0 s_1 \dots s_6]$

$\bar{s} = \bar{v} H^T$; where superscript T stands for matrix transpose.

The matrix H is called the parity check matrix and in the above example, it has size 7×16 .

$$H = \begin{bmatrix} 0 & 1 & 7 & 3 & 4 & 5 & 6 & 7 & 8 & 9 & 10 & 11 & 12 & 13 & 14 & 15 \\ 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 0 & 1 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 1 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 & 0 & 1 & 0 & 1 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \end{bmatrix} \quad 7 \times 16$$

notice that k is the length of the message bits, n is the length of code word,

then $G_{k \times n}$ and $H_{\substack{(n-k) \times n \\ \text{redundant bits}}}$

Do Example 4.2.2

4.3 Linear Block Codes

- An (n, k) linear block code encodes k -bit message vectors into n -bit code vectors. The code rate is $R = \frac{k}{n}$.
- In linear codes, the sum of any two codewords is a codeword $\Rightarrow \bar{c}_1 + \bar{c}_2 \in C$
- zero vector ($\bar{0}$) must be a codeword.
- We can define our codewords using a generator matrix G of size $k \times n$.

$$\bar{c} = \bar{m} G \quad ; \quad \text{where } \begin{array}{l} \bar{c} \text{ is } 1 \times n \text{ codeword} \\ \bar{m} \text{ is } 1 \times k \text{ message} \end{array}$$

Let the rows of G be $\bar{g}_0, \bar{g}_1, \dots, \bar{g}_{k-1}$

$$\text{then } \bar{c} = \bar{m} \begin{bmatrix} \bar{g}_0 \\ \bar{g}_1 \\ \vdots \\ \bar{g}_{k-1} \end{bmatrix} = m_0 \bar{g}_0 + m_1 \bar{g}_1 + \dots + m_{k-1} \bar{g}_{k-1}$$

- The number of codewords $|C| = 2^k$, since there are one 2^k distinct messages.

- The set of vectors $\{\bar{g}_i\}$ are linearly independent since we must have a set of unique codewords.

- linearly independent vectors mean that no vector \bar{g}_i can be expressed as a linear combination of the other vectors.

- These vectors are called basis vectors of the vector space C .

- The dimension of this vector space is the number of basis vectors which are k .

- $\bar{g}_i \in C \Rightarrow$ the rows of G are all legal codewords.

4.3.2

— Hamming weight of a vector is the number of nonzero elements in that vector

example $\bar{c} = [1001101] \Rightarrow w_H(\bar{c}) = 4$

— Hamming distance between two vectors is equal to the Hamming weight of their sum.

$$d_H(\bar{c}_1, \bar{c}_2) = w_H(\bar{c}_1 + \bar{c}_2)$$

— The minimum Hamming distance of a linear block code is equal to the minimum Hamming weight of the nonzero code vectors.

— Since each $\bar{g}_i \in C$, we must have $w_H(\bar{g}_i) \geq d_{\min}$.
this is a necessary condition but not sufficient.

Therefore, if ~~the~~ Hamming weight of one of the rows of G is less than d_{\min} , $\Rightarrow d_{\min}$ is not correct or G is not correct.

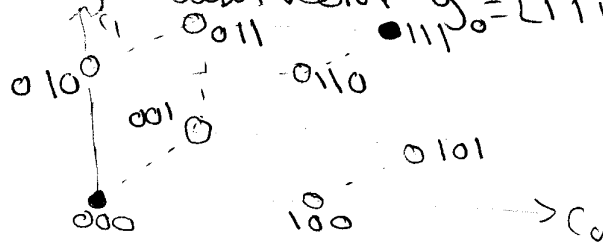
4.4 Decoding Linear Block codes

At the receiver,

$$\bar{v} = \bar{c} + \bar{e} ; \bar{v} \in V$$

and \bar{e} is the error vector of n bits.

Consider the $(3,1)$ repetition code, this is a 1-dimensional code with a single basis vector $\bar{g}_0 = [111]$.



So, we have only two valid codewords $(000, 111)$,

but the received vector \bar{v} , could be any one of these.

Def: Hamming Sphere

Hamming sphere of radius t is the set of all possible vectors \bar{v} that are at a Hamming distance less than or equal to t from a valid codeword.

$$P(n, t) = \sum_{j=0}^t \binom{n}{j}$$

Def: Hamming Bound

$$n - k = r \geq \log_2 (P(n, t))$$

this is a lower limit on the number of redundant bits.

Def: Perfect codes

A code that satisfies the Hamming bound with equality is called a "perfect code".

Minimum Distance Decoding

Lecture 15 P.2

The received vector $\bar{v} = \bar{c} + \bar{e}$, the decoder selects the codeword \bar{c} that minimizes the Hamming distance $d_H(\bar{v}, \bar{c})$.

Complete Decoders

Decoders that select \bar{c} which produce the minimum $d_H(\bar{v}, \bar{c})$

Bounded-distance Decoders

If $d_H(\bar{v}, \bar{c}) \leq t \Rightarrow$ select $\min d_H(\bar{v}, \bar{c})$
if $d_H(\bar{v}, \bar{c}) > t \Rightarrow$ Declare a decoder failure, "Error Detection"

Standard array decoders

The simplest, but most expensive strategy for implementing error correction is to simply look up the codeword (\bar{c}) in a decoding table that contains all possible \bar{v} . The table size grows exponentially with codeword length.

The lookup table will have 2^k columns and $2^{n-k} = 2^r$ rows.

See Example 4.4.1 in Page 137.

Syndrome Decoders and the Parity-Check Theorem

Lecture 15 P. 3

— The standard-array method becomes impractical with increasing block length.

— Syndrome decoders are more efficient and mostly used with Linear Block codes.

— The rows of the code's generator matrix G are basis vectors in the code space C . It is always possible to find basis vectors that give us a systematic code.

$$\Rightarrow G = \left[\begin{array}{c|c} P_{k \times r} & I_{k \times k} \end{array} \right]_{k \times n} \quad \left. \begin{array}{l} \\ \\ \end{array} \right\} r = n - k$$

where $I_{k \times k}$ is the $k \times k$ identity matrix and $P_{k \times r}$ is a $k \times r$ parity-bit generator.

The systematic code will be in this form

$$\bar{c} = \left[\underbrace{c_0 c_1 \dots c_{r-1}}_{\text{parity bits}} \mid \underbrace{m_0 m_1 \dots m_{k-1}}_{\text{information bits}} \right]$$

— At the receiver, the parity-check matrix is

$$H = \left[\begin{array}{c|c} I_{r \times r} & -P^T \end{array} \right]_{r \times n}$$

\Rightarrow For any code vector,

$$\bar{c} H^T = \bar{m} G H^T = \bar{m} \left[\begin{array}{c|c} P_{k \times r} & I_{k \times k} \end{array} \right] \begin{bmatrix} I_{r \times r} \\ -P_{k \times r} \end{bmatrix}$$

$$= \bar{m} \left[\begin{array}{c} P_{k \times r} - P_{k \times r} \end{array} \right] = \bar{0}$$

\circ Multiplying valid code vectors with H^T result in zero vector $\bar{0}$.

$$\boxed{\bar{c} H^T = \bar{0}}$$

— The syndrome vector is $\bar{s} = \{s_0, s_1, \dots, s_{r-1}\}$ Lecture 15 1.1

$$\begin{aligned}\bar{s} &= \bar{r} H^T \\ &= (\bar{c} + \bar{e}) H^T = \bar{c} H^T + \bar{e} H^T \\ &= \bar{0} + \bar{e} H^T = \bar{e} H^T\end{aligned}$$

∴ $\bar{s} = \bar{e} H^T$ a function only of the error vector.

⇒ \bar{s} is zero, if and only if \bar{e} is a valid code vector.

So \bar{s} is zero, $\left\{ \begin{array}{l} \text{if there are no errors} \Rightarrow \bar{e} = \bar{0} \\ \text{or if the errors make} \\ \bar{e} = \bar{c} \Rightarrow \text{valid code word} \end{array} \right.$

otherwise, $\bar{s} \neq \bar{0}$ and we have a detectable error. So, the decoder will either detect it or correct it.

— Also, notice that

$$G H^T = 0 \Rightarrow \text{the rows of } H \text{ is orthogonal to the rows of } G.$$

⇒ the rows of H must be also linearly independent

— The possible syndrome vectors are in one-to-one correspondence with the error patterns in the standard array.

Therefore, a particular syndrome uniquely identifies a particular error pattern.

— Thus, instead of storing the standard array with 2^k columns and 2^r rows, it is sufficient to store in a table the $2^r - 1$ nonzero error patterns corresponding to the $2^r - 1$ possible nonzero syndromes.

— Error Correction using the syndrome table.

For complete Decoders $\left\{ \begin{array}{l} \text{— If the syndrome is zero} \Rightarrow \text{no error.} \end{array} \right.$

$\left\{ \begin{array}{l} \text{— If the syndrome is nonzero} \Rightarrow \text{look up the error pattern for this syndrome} \end{array} \right.$

⇒ Add it to the received vector

For Bounded-distance decoders

- We store only those error patterns which satisfy our distance criterion.
- If we get a nonzero syndrome for which there is no table entry, we declare a decoder failure.