

KING FAHD UNIVERSITY OF PETROLEUM & MINERALS

ELECTRICAL ENGINEERING DEPARTMENT

FIRST SEMESTER 2014-2015 (S141)

Course Title:	Electronics II
Course Number:	EE 303

Exam Type:	Major Exam I
Date:	Sunday Oct. 19 th , 2014
Time:	5:30PM-7:00PM

Student Name: Key

Student ID: _____

Section # 5

GRADING		
Question 1	45	
Question 2	30	
Question 3	25	
Total:	100	

Show all your work and results. Do not give more than one answer otherwise the wrong one will be considered.

Question No.1 [45 points]

A. For the MOS pair shown below, assume $k'(W/L) = 4 \text{ mA/V}^2$, and $R_D = 3 \text{ k}\Omega$, , assume the current source output resistance (r_{o4}) of $50\text{k}\Omega$, and neglect channel-length modulation of Q1 &Q2.

- Find the single ended differential mode gain $A_{d1} = (V_{o1}/V_{id})$ where $V_{id} = V_{i1} - V_{i2}$. (10 marks)
- Find the single ended common mode gain $A_{cm1} = (V_{o1}/V_{cm})$. (10 marks)
- Find the single ended common mode rejection ratio CMRR. (5 marks)

$$i) i_d = \frac{N_1 - N_2}{\frac{2}{g_m}} = \frac{1}{2} g_m N_{id} \quad 4$$

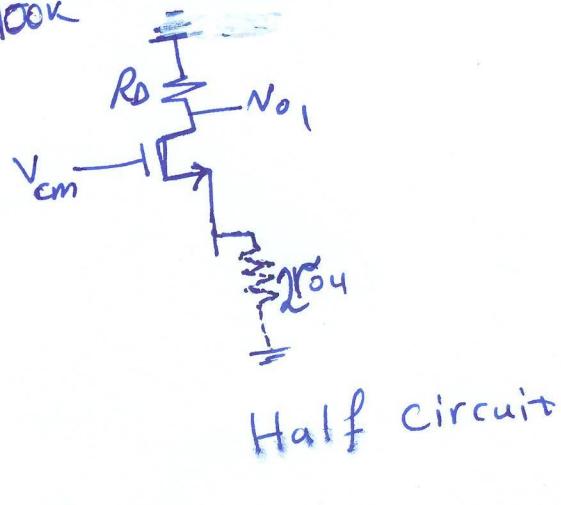
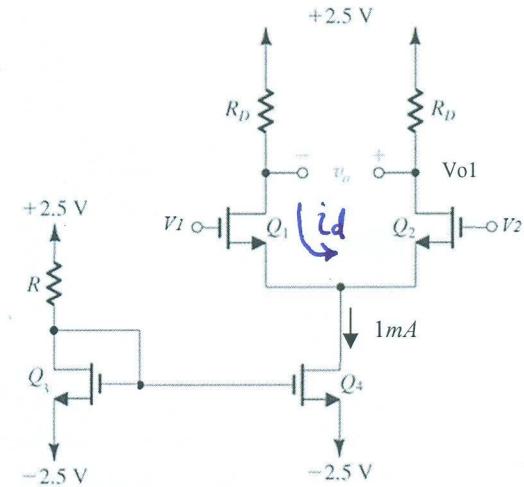
$$N_{o1} = i_d R_D \Rightarrow A_{d1} = \frac{N_{o1}}{N_{id}} = \frac{1}{2} g_m R_D \quad 4$$

$$A_{d1} = \frac{1}{2} \sqrt{2 \times 4 \times 0.5} \times 3 \text{ k} = \underline{\underline{3 \text{ V/V}}} \quad 2$$

$$ii) A_{cm1} = \frac{N_{o1}}{N_{cm}} = - \frac{R_D}{\frac{1}{g_m} + 2r_{o4}} = - \frac{3 \text{ k}}{0.5 \text{ k} + 100 \text{ k}} \quad 6$$

$$\approx -0.03 \quad \underline{\underline{V/V}} \quad 4$$

$$iii) CMRR = \left| \frac{A_{d1}}{A_{cm1}} \right| = \frac{3}{0.03} \quad 5$$



B. For the BJT pair shown below, $\beta=100$ & $\alpha \approx 1$, ignore r_o ,

- Find the differential mode gain $A_d = (v_o/v_{id})$. [10 points]
- Find the differential input resistance R_{id} . [10 points]

i

$$i_d = \frac{V_{id}}{2r_e + 200 + 200} \quad 4$$

$$r_e = \frac{V_T}{I_E} = \frac{25m}{0.25m} = 100\Omega \quad 2$$

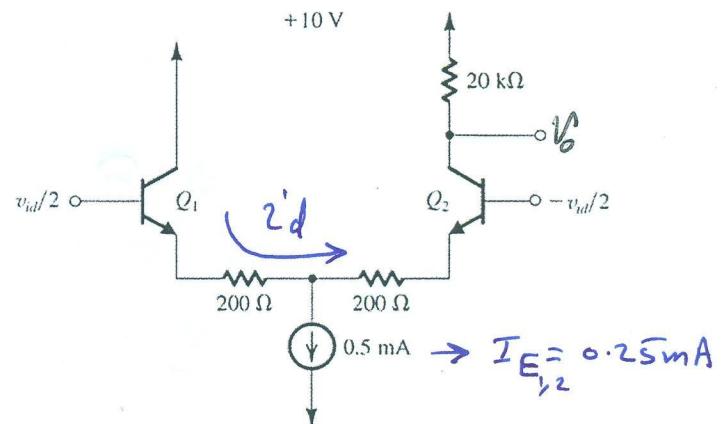
$$N_o = i_d \times 20k = \frac{i_d}{600} \times 20k \quad 2$$

$$A_d = \frac{N_o}{V_{id}} = 33.3 \quad 2$$

ii

$$R_{id} = 2 \times r_\pi + (\beta + 1)(200 + 200) \quad 8$$

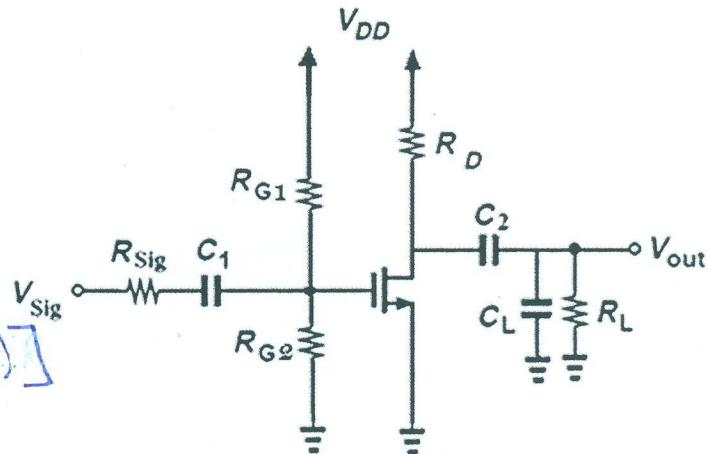
$$= 2 \times (101)(100) + (101)(400) = 60600\Omega \quad 2$$



Question No. 2 [30 points]

In the circuit shown, $g_m = 1 \text{ ms}$, $C_{gs} = 10 \text{ pF}$, $C_{gd} = 1 \text{ pF}$, $R_{sig} = 10 \text{ k}\Omega$, $R_{G1} = 2 \text{ M}\Omega$, $R_{G2} = 2 \text{ M}\Omega$, $R_D = 18 \text{k}\Omega$, $C_L = 50 \text{ pF}$, and $R_L = 18 \text{ k}\Omega$.

- a- Drive the expressions for all high frequency poles. [12 points]
- b- Find the high corner frequency ω_H . [8 points]
- c- Find the Midband voltage gain $A_M = V_{out}/V_{sig}$. [10 points]



a)

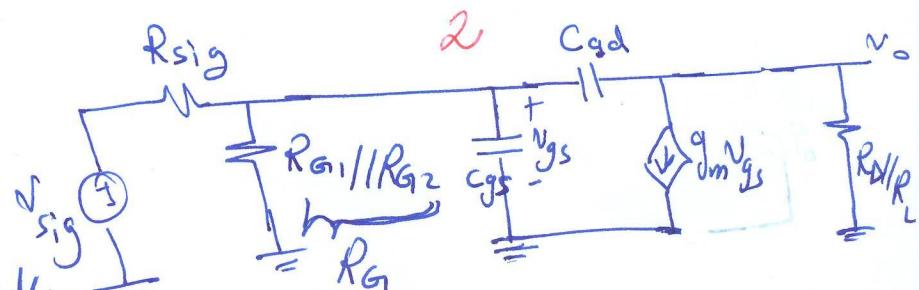
$$C_{eq} = C_{gs} + C_1$$

$$C_1 = C_{gd}(1 - K) = C_{gd} \left[1 + g_m (R_{D} // R_L) \right] \\ = 1 \text{ pF} [1 + 1 \text{ m} \times 9 \text{ K}] = 10 \text{ pF}$$

$$C_{eq} = 10 \text{ pF} + 10 \text{ pF} = 20 \text{ pF}$$

$$\omega_{H1} = \frac{1}{C_{eq} [R_{sig} // R_{G1}]} = 5 \text{ rad/sec}$$

$$\omega_{H1} = \frac{1}{20 \text{ pF} [10 \text{ k}\Omega // 1 \text{ M}\Omega]} \approx 5 \text{ rad/sec} \quad 4$$



$$\omega_{H2} = \frac{1}{(C_1 + C_2)(R_D // R_L)} = 2.174 \text{ rad/sec}$$

$$= \frac{1}{51.0 \text{ pF} \times 9 \text{ k}\Omega} = 2.174 \text{ rad/sec} \quad 4$$

$$b) \quad \omega_H = \frac{1}{\frac{1}{\omega_{H1}} + \frac{1}{\omega_{H2}}} = 1.52 \text{ rad/sec} \quad 8$$

$$c) \quad A_M = \frac{V_{out}}{V_{sig}} = \frac{R_D}{R_{sig} + R_G} (-g_m R_D // R_L) = -0.99 \times (-1 \text{ m} \times 9 \text{ K}) \\ = -8.91 \text{ V/V} \quad 10$$

Question No. 3 [25 points]

For the common base amplifier shown; assumes r_x and r_o are neglected:

a- Drive the expressions for all low frequency poles. [10 points]

b- Find an expression for the low corner frequency f_L (do not neglect any pole). [5 points]

c- Drive the expressions for the Midband voltage gain $A_M = V_o/V_{sig}$. [10 points]

a) Low freq. poles:

$$\omega_{L1} = \frac{1}{C_B R_{CB}}$$

$$R_{CB} = R_1 // R_2 // [\pi + (\beta + 1) R_{sig}]$$

$$\omega_{L2} = \frac{1}{C_C (R_C + R_L)}$$
(10)

b) $\omega_L \approx \omega_{L1} + \omega_{L2}$ 5

c) $A_M = \frac{R_C // R_L}{r_e + R_{sig}}$ 10

