

KING FAHD UNIVERSITY OF PETROLEUM & MINERALS
ELECTRICAL ENGINEERING DEPARTMENT
SECOND SEMESTER 2010-2011 (102)



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|-----------------------|------------------------------------|
| Course Title: | Communication Engineering I |
| Course Number: | EE 370 |

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| Exam Type: | MAJOR EXAM I |
| Date: | March 23, 2011 |
| Time: | 7:00 pm – 8:30 pm (1 & 1/2 hours) |

Student Name: _____ **Key** _____

Student ID: _____ **00000** _____

Section: _____

Serial Number: _____

| GRADING | | |
|-------------------|-----------|--|
| Question 1 | 10 | |
| Question 2 | 10 | |
| Question 3 | 20 | |
| Question 4 | 20 | |
| Total: | 60 | |

Be neat, organized, and show all your work and results.

Question 1:

Mark the following clearly as true (T) or false (F).

| | | |
|----|---|-------|
| 1 | DSBSC has a power efficiency of about 67% which makes it more power efficient compared to AM signal having a power efficiency of 33% at best. | (F) |
| 2 | AM signal can be demodulated coherently | (T) |
| 3 | When using switching modulators for DSBSC, it is necessary to use double balanced modulators. | (F) |
| 4 | In compact trigonometric Fourier series, $(C_n = \sqrt{a_n^2 + b_n^2})$ is complex and it contains the amplitude and phase information of the frequency spectra. | (F) |
| 5 | The amplitude spectrum of a real signal is even and the phase spectrum is odd. | (T) |
| 6 | Both AM and DSBSC modulations need twice the bandwidth of the modulating signal and a carrier frequency of at least twice the bandwidth of the modulating signal. | (F) |
| 7 | Distortionless systems have constant amplitude spectrum and exponential phase spectrum. | (F) |
| 8 | The time constant, RC, of the low-pass filter for the envelope detector depends on the value of the modulation index. | (T) |
| 9 | In QAM demodulation, phase mismatch is less damaging than frequency mismatch | (T) |
| 10 | Band-limited signals have infinite bandwidth. | (F) |

Question 2:

- a) Find the complex (exponential) Fourier series of $x(t) = \cos(2000\pi t) + \sin^2(2000\pi t)$ (4 marks)
- b) Plot the amplitude spectrum and phase spectrum (two sided) of the signal $x(t)$. (4 marks)
- c) Show whether $x(t)$ is a power or energy signal and find its corresponding power or energy if possible. (2 marks)

$$\begin{aligned}
 a) \quad x(t) &= \cos(2000\pi t) + \frac{1}{2} [1 - \cos 4000\pi t] \\
 &= \frac{1}{2} + \cos(2000\pi t) - \frac{1}{2} \cos 4000\pi t \\
 &= \frac{1}{2} + \frac{1}{2} [e^{j2000\pi t} + e^{-j2000\pi t}] - \frac{1}{4} [e^{j4000\pi t} + e^{-j4000\pi t}] \\
 &= \frac{1}{2} + \frac{1}{2} e^{j2000\pi t} + \frac{1}{2} e^{-j2000\pi t} - \frac{1}{4} e^{j4000\pi t} - \frac{1}{4} e^{-j4000\pi t}
 \end{aligned}$$

note that $x(t) = \sum_{n=-\infty}^{\infty} D_n e^{jn\omega_0 t}$

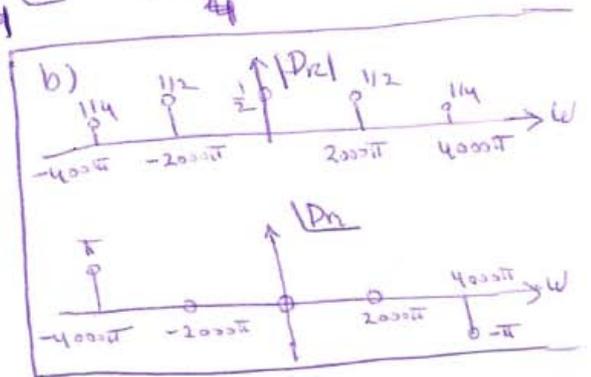
$D_0 = \frac{1}{2} \text{ at } \omega = n\omega_0 = 0$

$D_1 = \frac{1}{2} \text{ at } \omega = n\omega_0 = 2000\pi$

$D_{-1} = \frac{1}{2} \text{ at } \omega = n\omega_0 = -2000\pi$

$D_2 = \frac{-1}{4} = \frac{1}{4} \angle -\pi \text{ at } \omega = n\omega_0 = 4000\pi$

$D_{-2} = \frac{-1}{4} = \frac{1}{4} \angle \pi \text{ at } \omega = n\omega_0 = -4000\pi$



c) $x(t)$ is periodic \Rightarrow it is a power signal

Using Parseval's theorem

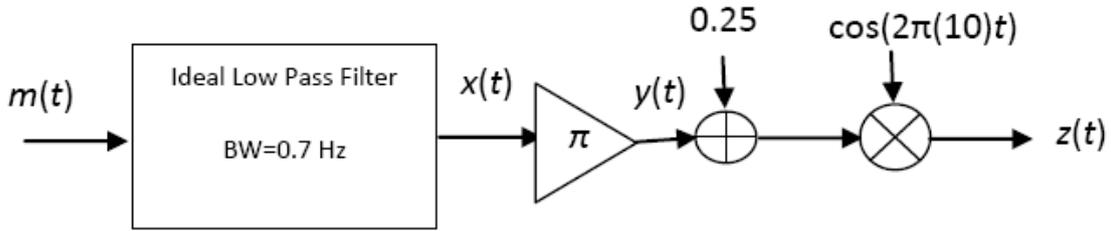
$$\begin{aligned}
 P_x &= \sum_{n=-\infty}^{\infty} |D_n|^2 = \left(\frac{1}{2}\right)^2 + \left(\frac{1}{2}\right)^2 + \left(\frac{1}{2}\right)^2 + \left(\frac{1}{4}\right)^2 + \left(\frac{1}{4}\right)^2 \\
 &= \frac{7}{8} = 0.875 \text{ watt}
 \end{aligned}$$

You can use also

$$\begin{aligned}
 P_x &= C_0^2 + \frac{1}{2} \sum_{n=1}^{\infty} c_n^2 = \left(\frac{1}{2}\right)^2 + \frac{1}{2} (1)^2 + \frac{1}{2} \left(\frac{1}{2}\right)^2 \\
 &= \frac{7}{8} = 0.875 \text{ watt}
 \end{aligned}$$

Question 3:

The system below, takes the input signal, low-pass filter it with cut off frequency of 0.7 Hz, then amplify it with a gain equals to π , then add a constant of 0.25, and finally multiply the resultant with a carrier of 10Hz



The input signal, $m(t)$, is periodic and can be represented by its Fourier series expansion as:

$$m(t) = \sum_{n=1}^{\infty} \frac{2}{n\pi} \cos(n\pi t + \theta_n), \quad \text{where } \theta_n = \begin{cases} -\frac{\pi}{2} & n \text{ odd} \\ +\frac{\pi}{2} & n \text{ even} \end{cases}$$

- a. Find the percentage of power at the output of the filter compared to the input power. Hint: you will need to find the power of $m(t)$ and $x(t)$ and compare them.

$$P_{g(t)} = \frac{1}{T_0} \int_{t_0}^{t_0+T_0} |g(t)|^2 dt$$

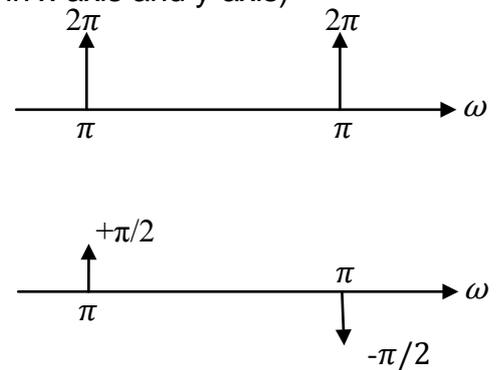
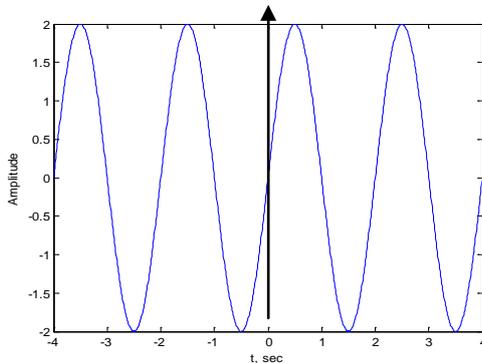
$$P_{m(t)} = \frac{1}{2} \int_{-1}^1 t^2 dt = \frac{1}{2} \frac{1}{3} [(1)^3 - (-1)^3] = \frac{1}{3}$$

After the low-pass filter only the first harmonic will pass, Note that $T_0=2 \rightarrow f_0=0.5 \text{ Hz}$

The first harmonic is given to be $x(t) = \frac{2}{\pi} \cos(\pi t - \frac{\pi}{2}) = \frac{2}{\pi} \sin(\pi t)$. Power of sinusoidal is $\frac{(\frac{2}{\pi})^2}{2} = \frac{2}{\pi^2} = 0.2026$

Power ratio = $(0.2026)/(1/3) = 0.6079 \approx 60.8\% \dots$ (5 points) Sketch the signal $y(t)$ in time domain and its magnitude and phase spectra. (show all values in x-axis and y-axis)

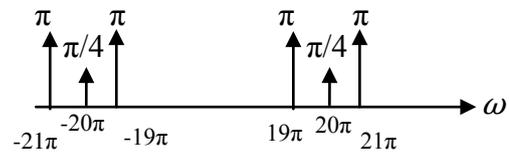
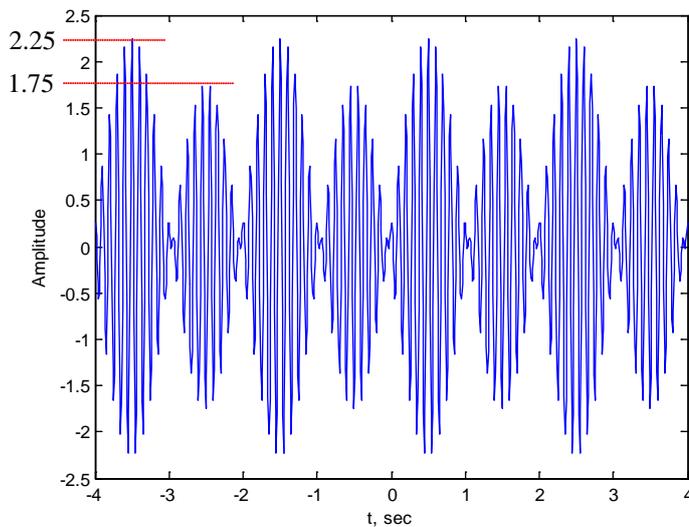
$$y(t) = 2\sin(\pi t)$$



(3 points, note that for the sin is 0 at t=0)

If you could not solve the above assume $y(t) = \cos(\pi)t$

b. Sketch $z(t)$ and its magnitude spectrum. (show all important values on the sketch)



(4 points)

c. Notice that $z(t)$ is the AM modulated signal for $y(t)$, find the values of the modulation index and the power efficiency? Comment on the value of the efficiency.

$$\mu = \frac{m_p}{A} = \frac{2}{0.25} = 8$$

$$\text{Power efficiency} = \frac{\tilde{m}^2}{\tilde{m}^2 + A^2} = \frac{2^2/2}{\frac{2^2}{2} + (0.25)^2} = 0.977, 97.7\%. \text{ We can also use the other equation}$$

because it is single tone

$$\text{Power efficiency} = \frac{\mu^2}{\mu^2 + 2} = \frac{64}{64 + 2} = 0.977$$

The value of the efficiency is very high and it is expected not to exceed 33.33 % for single tone if a non coherent detector is to be used. This system is over-modulated and cannot be recovered non coherently

(4 points)

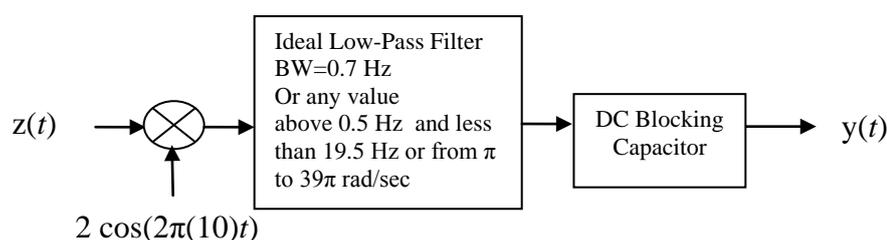
d. Can $y(t)$ be recovered from $z(t)$ using a non-coherent detector? Why?

No, because the system is over-modulated. We did not add enough carrier. We still cannot distinguish the negative from the positive side for the message.

(2 points)

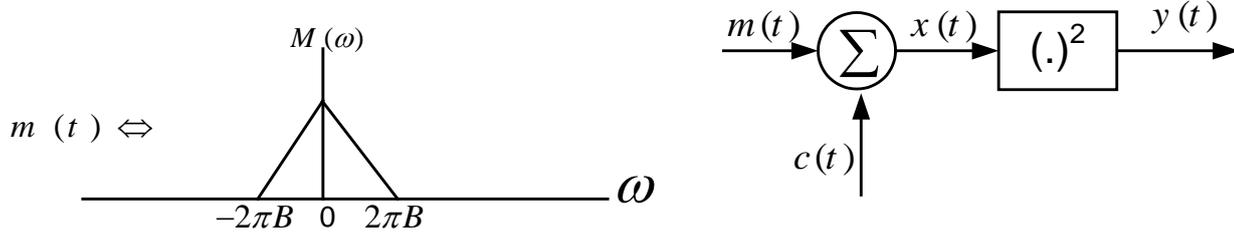
e. Sketch a system that can recover $y(t)$ from $z(t)$. (specify all possible values for the filter)

(2 points)



Question 4:

For the system shown, let the signal $c(t) = \cos \omega_c t$ and let $m(t)$ have the following spectrum:



- a) Find the time domain signal $y(t)$ in terms of $m(t)$ and sketch its amplitude spectrum. (6 marks)
- b) What must you do to $y(t)$ to get DSBSC modulated signal out of it. (3 marks)
- c) Assume you used instead of $m(t)$ the modulated signal $m(t)c(t)$, find the time domain signal $y(t)$ in terms of $m(t)$. (5 marks)
- d) From your result in part c, can this system be used to demodulate DSBSC signal? If yes, how; if not, why. (6 marks)

assume $\omega_c \gg 2\pi B$

a) $y(t) = [m(t) + c(t)]^2$

$$= m^2(t) + c^2(t) + 2m(t)c(t)$$

$$= m^2(t) + \frac{1}{2} + \frac{1}{2} \cos(2\omega_c t) + 2m(t) \cos \omega_c t$$

b) use a bandpass filter of centre frequency ω_c & BW of $4\pi B$.

c) $y(t) = [m(t)c(t) + c(t)]^2$

$$= m^2(t)c^2(t) + c^2(t) + 2m(t)c^2(t)$$

$$= m^2(t) \left[\frac{1 + \cos 2\omega_c t}{2} \right] + \frac{1 + \cos 2\omega_c t}{2} + m(t) + m(t) \cos(2\omega_c t)$$

$$= \frac{1}{2} m^2(t) + \frac{1}{2} m^2(t) \cos 2\omega_c t + \frac{1}{2} + \frac{1}{2} \cos 2\omega_c t + m(t) + m(t) \cos 2\omega_c t$$

d) at the baseband we have $\frac{1}{2} m^2(t) + \frac{1}{2} + m(t)$

So if we use LPF of BW $2\pi B$ we will get $\frac{1}{2} + m(t) + \frac{1}{2} m^2(t)$

we can get $m(t)$ if $m(t) \ll 1$

So, $\frac{m^2(t)}{2}$ is negligible & using capacitor, we can get rid of the DC.