

EE 205 Dr. A. Zidouri

Electric Circuits II

Two-Port Circuits
Interconnected Two-Port Circuits

Lecture #44

EE 205 Dr. A. Zidouri

The material to be covered in this lecture is as follows:

- Terminated Two-Port circuit
- Terminal Behavior
- The Six characteristics of the terminated two-port circuit in terms of z parameters
- Interconnected Two-Port Circuits

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After finishing this lecture you should be able to:

- Analyze The Terminated Two-Port circuit
- Determine The characteristics of the terminated circuit in terms of z parameters
- Recognize the different Interconnection of the Two-Port Circuits
- Analyze The Cascade Connection

Terminated Two-Port Circuit

- ✚ The Circuit is driven at port 1 and loaded at port 2
- ✚ A typically terminated two-port model is shown in Fig. 44-1

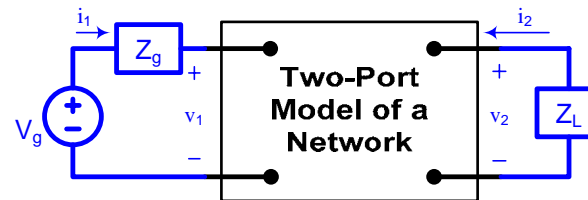


Fig. 44-1 A Terminated Two-Port Model

- ✚ Z_g represents the internal impedance of the source
- ✚ Z_L represents the Load impedance
- ✚ V_g represents the internal voltage of the source
- ✚ Analysis of this circuit involves expressing the terminal currents and voltages as function of V_g , Z_L , and Z_g .

Terminal Behavior:

Six characteristics of the terminated two-port circuit define its terminal behavior.

1. The input impedance $Z_{in} = \frac{V_1}{I_1}$ or admittance $Y_{in} = \frac{I_1}{V_1}$
2. The output current I_2
3. Thevenin voltage and impedance (V_{Th}, Z_{Th}) with respect to port 2

4. The current gain $\frac{I_2}{I_1}$

5. The voltage gain $\frac{V_2}{V_1}$

6. The current gain $\frac{V_2}{V_g}$

The six characteristics in Terms of the z Parameters:

- ✚ We develop the expressions using the z-parameters to model the two-port portion of the circuit.
- ✚ Expressions involving other parameters **y**, **a**, **b**, **h** and **g** can be found in tables in text books.
- ✚ The derivation of any one of the desired expressions involves the algebraic manipulation of the two-port equations along with the two constraint equations.
- ✚ These four equations using z parameters are:

$$\text{i. } V_1 = z_{11}I_1 + z_{12}I_2 \quad (44-1)$$

$$\text{ii. } V_2 = z_{21}I_1 + z_{22}I_2 \quad (44-2)$$

$$\text{iii. } V_1 = V_g - I_1 Z_g \quad (44-3)$$

$$\text{iv. } V_2 = -I_2 Z_L \quad (44-4)$$

To find the input impedance $Z_{in} = \frac{V_1}{I_1}$ we proceed as follows:

In (44-2) we replace V_2 from (44-4) we solve for I_2 we get:
$$I_2 = \frac{-z_{21}I_1}{Z_L + z_{22}} \quad (44-5)$$

We then substitute in (44-1) and solve for Z_{in} we get:
$$Z_{in} = z_{11} - \frac{z_{12}z_{21}}{z_{22} + Z_L} \quad (44-6)$$

To find I_2 we first solve (44-5) for I_1 after replacing V_1 with the RHS of (44-3) the result is:

$$I_1 = \frac{V_g - z_{12}I_2}{Z_g + z_{11}} \quad (44-7)$$

We now substitute (44-7) into (44-5) and solve for I_2 :
$$I_2 = \frac{-z_{21}V_g}{(Z_g + z_{11})(Z_L + z_{22}) - z_{12}z_{21}} \quad (44-8)$$

The Thevenin voltage with respect to port 2 equals V_2 when $I_2 = 0$. Therefore:

$$V_2|_{I_2=0} = z_{21}I_1 = z_{21} \frac{V_1}{Z_g + z_{11}} \quad (44-9)$$

But $V_1 = V_g - I_1 Z_g$, and $I_1 = \frac{V_g}{Z_g + z_{11}}$ when $I_2=0$; therefore by substitution into (44-9) the open circuit

value of V_2 is:

$$V_2 \Big|_{I_2=0} = V_{Th} = \frac{z_{21}}{Z_g + z_{11}} V_g \quad (44-10)$$

The Thevenin or output impedance is the ratio $Z_{Th} = \frac{V_2}{I_2}$ when V_g is zero (short circuit). Thus (44-3)

$$\text{becomes } V_1 = -I_1 Z_g \quad (44-11)$$

$$\text{Substituting gives } I_1 = \frac{-z_{12} I_2}{Z_g + z_{11}} \quad (44-12)$$

$$\text{Therefore } \frac{V_2}{I_2} \Big|_{V_g=0} = Z_{Th} = z_{22} - \frac{z_{12} z_{21}}{Z_g + z_{11}} \quad (44-13)$$

$$\text{The current gain comes directly from (44-5): } \frac{I_2}{I_1} = \frac{-z_{21}}{Z_L + z_{22}} \quad (44-14)$$

To derive the voltage gain $\frac{V_2}{V_1}$ we replace I_2 in (44-2) with its value from (44-4):

$$V_2 = z_{21} I_1 + z_{22} \left(\frac{-V_2}{Z_L} \right) \quad (44-15)$$

Next we solve (44-2) for I_1 in terms of V_1 and V_2 :

$$z_{11}I_1 = V_1 - z_{12} \left(\frac{-V_2}{Z_L} \right)$$

Or

$$I_1 = \frac{V_1}{z_{11}} + \frac{z_{12}V_2}{z_{11}Z_L} \quad (44-16)$$

Replace I_1 in (44-15) and solve:

$$\frac{V_2}{V_1} = \frac{z_{21}Z_L}{z_{11}Z_L + z_{11}z_{22} - z_{12}z_{21}} = \frac{z_{21}Z_L}{z_{11}Z_L + \Delta z} \quad (44-17)$$

To derive the voltage gain $\frac{V_2}{V_g}$ we first find I_1 in terms of V_1 and V_2 by combining

(44-1), (44-3) and (44-4):

$$I_1 = \frac{V_g}{z_{11} + Z_g} + \frac{z_{12}V_2}{Z_L(z_{11} + Z_g)} \quad (44-18)$$

Then using (44-3) and (44-18) with (44-2) we derive an expression involving only V_2 and V_g :

$$V_2 = \frac{z_{21}z_{12}V_2}{Z_L(z_{11} + Z_g)} + \frac{z_{21}V_g}{z_{11} + Z_g} - \frac{z_{22}}{Z_L}V_2 \quad (44-19)$$

After manipulation we get:

$$\frac{V_2}{V_g} = \frac{z_{21}Z_L}{(z_{11} + Z_g)(z_{22} + Z_L) - z_{12}z_{21}} \quad (44-20)$$

Example 44-1

The two-port circuit of Fig. 44-2 is described in terms of its b parameters, the values of which are:

$b_{11} = -20$, $b_{12} = -3k\Omega$, $b_{21} = -2mS$, and $b_{22} = -0.2$

- Find the phasor voltage V_2
- Find the average power delivered to the $5k\Omega$ load
- Find the average power delivered to the input port
- Find the load impedance for maximum average power transfer
- Find the maximum average power delivered to the load in (d)

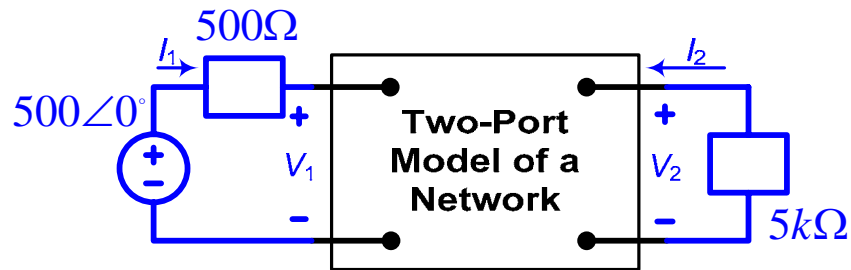


Fig. 44-2 The Circuit for Example 44-1

Solution:

- To find V_2 , we can either use (44-4) or from the voltage gain in (44-20), using the latter approach and conversion tables:

$$\Delta b = (-20)(-0.2) - (-3000)(-0.002) = -2$$

$$\frac{V_2}{V_g} = \frac{z_{21}Z_L}{(z_{11} + Z_g)(z_{22} + Z_L) - z_{12}z_{21}}$$

$$= \frac{\Delta b Z_L}{b_{12} + b_{11}Z_g + b_{22}Z_L + b_{21}Z_gZ_L} = \frac{(-2)(5000)}{-3000 + (-20)(500) + (-0.2)(5000) + (-0.002)(500)(5000)}$$

$$= \frac{10}{19} \text{ Then, } V_2 = \frac{10}{19} 500 = 263.16 \angle 0^\circ \text{ V}$$

b) The average power delivered to the 5000Ω load is

$$P_2 = \frac{263.16^2}{2 \times 5000} = 6.93 \text{ W}$$

c) To find average power delivered to the input port, we first find the input impedance Z_{in} ,

$$\text{From tables, } Z_{in} = \frac{b_{22}Z_L + b_{12}}{b_{11} + b_{21}Z_L} = \frac{(-0.2)(5000) - 3000}{(-0.002)(5000) - 20} = 133.33\Omega$$

$$\text{Therefore } I_1 \text{ is } I_1 = \frac{500}{500 + 133.33} = 789.47 \text{ mA}$$

The average power delivered to the input port is $P_1 = \frac{0.78947^2}{2} 133.33 = 41.55W$

d) The total impedance for maximum power transfer Z_L is the conjugate of Thevenin impedance.

$$Z_{Th} = \frac{b_{11}Z_g + b_{12}}{b_{22} + b_{21}Z_g} = \frac{(-20)(500) - 3000}{(-0.002)(500) - 0.2} = 10833.33\Omega$$

Therefore $Z_L = Z_{Th}^* = 10833.33\Omega$

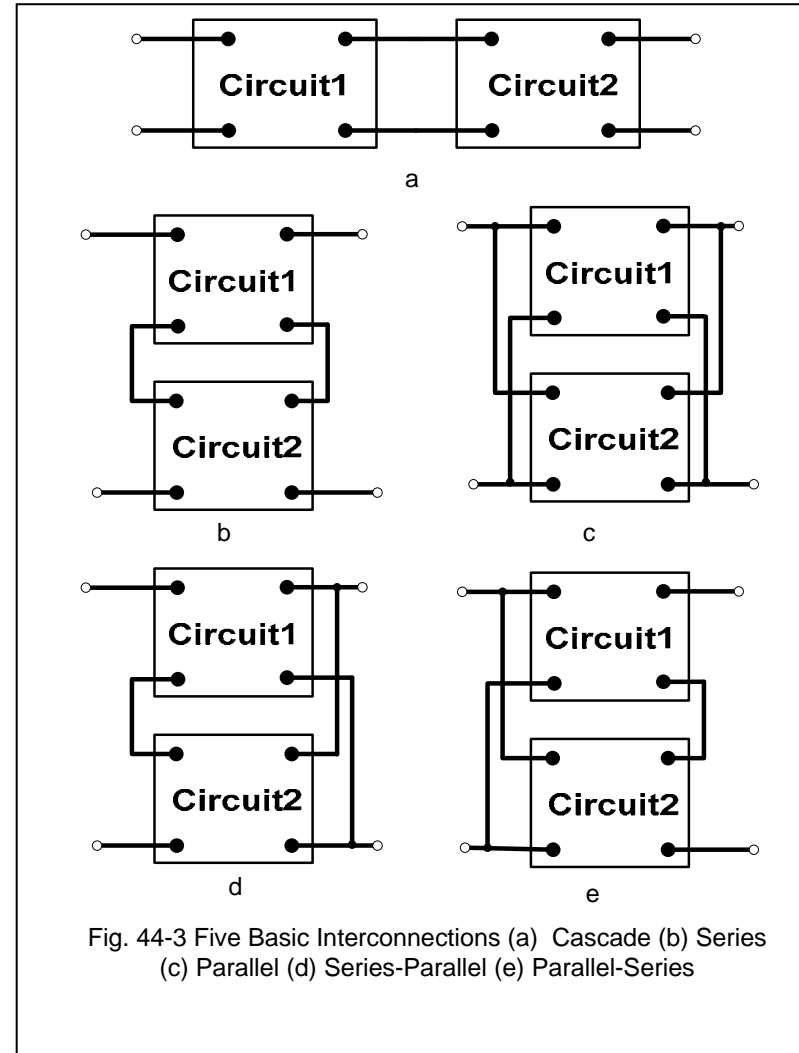
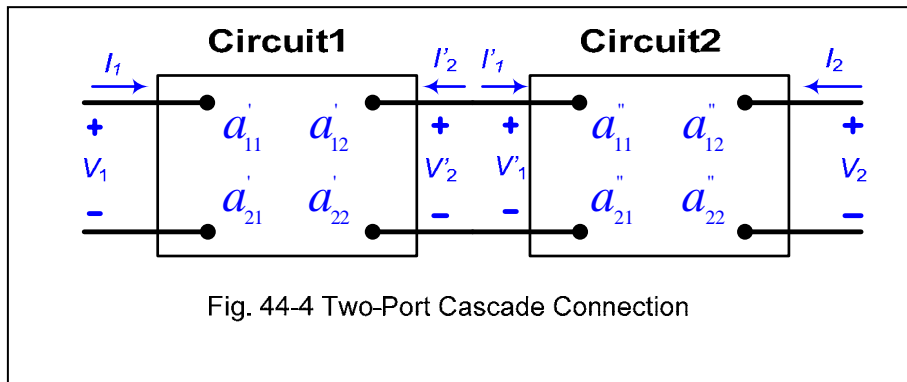
e) To find the maximum average power delivered to the load, we first find V_2 from the gain expression when $Z_L = 10833.33\Omega$

$$\frac{V_2}{V_1} = 0.8333 \text{ thus } V_2 = 0.8333 \times 500 = 416.67V \text{ and } P_{2\max} = \frac{1}{2} \frac{416.67^2}{10833.33} = 8.01W$$

Interconnected Two-Port Circuits

Two-port circuits may be interconnected in five ways:

- 1) In Cascade Fig. 44-3a
- 2) In Series Fig. 44-3b
- 3) In Parallel Fig. 44-3c
- 4) In Series-Parallel Fig. 44-3d
- 5) In Parallel-Series Fig. 44-3e



- ✚ We analyze only the Cascade connection because it occurs frequently in the modeling of large systems.
- ✚ The a-parameters are best suited for describing cascade connection
- ✚ We seek the pair of equations:

$$V_1 = a_{11}V_2 - a_{12}I_2 \quad (44-1)$$

$$I_1 = a_{21}V_2 - a_{22}I_2 \quad (44-2)$$

From Fig. 44-4 we have

$$V_1 = a'_{11}V'_2 - a'_{12}I'_2 \quad (44-3)$$

$$I_1 = a'_{21}V'_2 - a'_{22}I'_2 \quad (44-4)$$

From interconnection $V'_2 = V_1$ and $I'_2 = -I_1$, then substituting yields:

$$V_1 = a'_{11}V_1 - a'_{12}I_1 \quad (44-5)$$

$$I_1 = a'_{21}V_1 - a'_{22}I_1 \quad (44-6)$$

From the second circuit

$$V'_1 = a''_{11}V_2 - a''_{12}I_2 \quad (44-7)$$

$$I'_1 = a''_{21}V_2 - a''_{22}I_2 \quad (44-8)$$

By substitution we generate

$$V_1 = (a'_{11}a''_{11} + a'_{12}a''_{12})V_2 - (a'_{11}a''_{12} + a'_{12}a''_{22})I_2 \quad (44-9)$$

$$I_1 = (a'_{21}a''_{11} + a'_{22}a''_{21})V_2 - (a'_{21}a''_{12} + a'_{22}a''_{22})I_2 \quad (44-10)$$

By comparison we get the desired expressions:

$$a_{11} = a'_{11} a''_{11} + a'_{12} a''_{12} \quad (44-11)$$

$$a_{12} = a'_{11} a''_{12} + a'_{12} a''_{22} \quad (44-12)$$

$$a_{21} = a'_{21} a''_{11} + a'_{22} a''_{21} \quad (44-13)$$

$$a_{22} = a'_{21} a''_{12} + a'_{22} a''_{22} \quad (44-14)$$

Self Test 44:

Find the transmission parameters for the circuit in Fig. 44-5

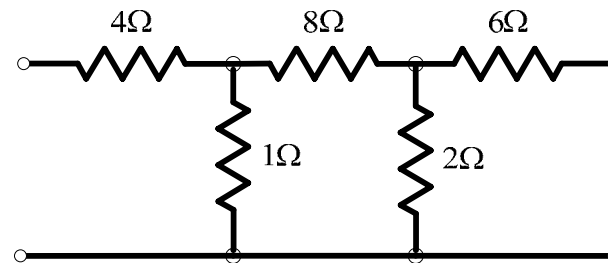


Fig. 44-5 The Circuit for Self Test 44

Answer:

$$[T] = [T_1][T_2] = \begin{bmatrix} 5 & 44 \\ 1 & 9 \end{bmatrix} \begin{bmatrix} 1 & 6 \\ 0.5 & 4 \end{bmatrix} = \begin{bmatrix} 27 & 206\Omega \\ 5.5S & 42 \end{bmatrix}$$

i.e. $a_{11} = 27$ $a_{12} = 206\Omega$ $a_{21} = 5.5S$ $a_{22} = 42$